Smart Grid Architceture

Akash K Singh, PhD

IBM Corporation Sacramento, USA

Abstract

At present, "Smart Grid" emerged to be one of the best advanced energy supply chain. This paper looks into the security system of Smart Grid via Smart Planet system. The scope focused onto information security criteria that impacts consumer trust and satisfaction. The importance of information security criteria is the main aspect perceived to impact customer trust towards the entire smart grid system. On one hand, it also focused on the selection of the model in developing information security criteria on smart grid.

In the power grid, efficient coordination among electricity generation, transmission, distribution and consumption processes call for integration of the advances in Information and Communication Technologies (ICT) to the physical components of the grid. The need for coordination and control becomes even more pronounced when the additional loads of the Plug- In Hybrid Electrical Vehicles (PHEVs) are considered. PHEVs are anticipated to be widely adopted in the following years, and this will increase the load on the power grid since the batteries of the PHEVs will be charged mostly from the grid supplied power. In this case, avoiding mismatch between generation and consumption is one aspect of the problem, whereas to avoid overloading the distribution system components, e.g. transformers, are another equally important challenge. In this paper, we consider an architecture where the status of the grid is monitored by the utility and translated into an amount of provisioned energy for each distribution system serviced by a substation. The substation employs an admission control mechanism for the PHEV charge demands based on the provisioned energy amount. We provide the theoretical analysis of this admission control scheme by calculating the blocking probability of the PHEV demands. We also propose a mechanism to reduce the load without increasing the blocking probability. We introduce an activity factor in the model and show that it can be used to reduce the load. We show by theoretical analysis and simulations that our PHEV admission control mechanism decreases the overall load in the system, and hence increases the resilience of the smart grid. Meanwhile, we show that load reduction can be implemented without increasing the blocking probability, thus customer satisfaction is not degraded.

I. INTRODUCTION- DESIGN METHODOLOGY IN A FEEDBACK CONTROL ARCHITECTURE

Our rule-based controller design methodology with a feedback control architecture is illustrated in Fig. 1. The model of the system to be controlled (rule-based power system model in this case) is represented by the block p($\cdot$) in Fig. 1. The rulebased controller h($\cdot$) is a feedback controller that acts on the load forecast $l(t)$, current environmental factors $z(t)$ such as the water level of the hydro generator and weather prediction, and the error signals $e(t)$, which are the difference between the actual load $y_{actual}(t)$ at time t and the output load $y_{out}(t)$ from the power system model for time t ($y_{actual}(t)$ and $y_{out}(t)$ are in the same domain). The rule-based controller consists of transition rules that map the inputs $l(t)$, $z(t)$ and $e(t)$ to the output $u(t)$ that then generates the power supply control schedule. The vector $u(t)$ represents the input to the power system model, or control action (e.g. to turn on a generator) of the power system at time t. We assume all the rules used in the power system model and the controller are written in Horn clauses. When a set of rules written in Horn clauses are sound and complete, a trim and minimal finite state Moore machine exists for the rule-based system according to the Theorem in [7] as follows:

Theorem 1: If a finite set of rules written in Horn clauses is sound and complete, then there exists a finite automaton that is trim and minimal.

The rule-based controller can then be represented as a 6- tuple Moore machine M=(Q, Σ, Y, δ, Λ, q_0), where Q is a set of states {q}, Σ is the input alphabet (a finite, non-empty set of symbols {σ}), Y is the set of output alphabet (a finite, nonempty set of symbols {υ}), δ is the state-transition function, that is $\delta : Q \times \Sigma \to Q$, Λ is the output function, that is $\Lambda : Q \to Y$, and q_0 is an initial state, an element of Q. We define each component of the Moore machine for the example rule-based controller in Section 3. Suppose we want

to minimize the cost of the difference between the actual load yactual(t) and the output load yout(t) from the power system model, and the cost of the control actions, which is a common criterion in practice. An optimal control problem can be formulated to solve for the set of operational rules that constitute the rule-based controller h(·) with given p(·), l(t), z(t) and yactual(t). A detailed formulation of the optimal control problem with a numerical example is given in Section 3.

A. Petri Net Modeling

The proposed modeling method includes two steps: first, the overall performance and cost of each IT infrastructure is determined based on scored performance and cost FoMs; then, it is determined that how much the suggested IT infrastructures satisfy and adapt to the given policy. Fig. 8 shows the designed Petri net for determining each alternative's performance based on its PFoMs (CAIDI, ASUI, and Rel.). As mentioned earlier, to be able to apply scored FoMs to the net as inputs, they should be quantized. In our application, there are three levels of quantization including Unusable (U), Moderate (M), and Excellent (E). These quantized scored FoMs form initial marking in a way that Unusable, Moderate, and Excellent values are modeled by one, two, and three number of tokens in corresponding places, respectively. The output of the net is the overall performance of the alternative which can be one of the three values of Poor, Moderate, or Excellent corresponding to places Pa, Ma, and Ea, respectively. As shown, we used a hierarchical algorithm in our modeling method, i.e. the input tokens in initial marking propagate in the net from left to right filling output places hierarchically until the net reaches its dead-end. To illustrate, the proposed net is the implementation of the look-up table of Table I extracted from experts' judgments; for instance, when the IT infrastructure in Unusable regarding one (or more) PFoM, disregarding values of other PFoMs, the overall performance of alternative is Poor. The modeling method is in a way that when the net reaches its dead-end, there is one token in only one of the output places Pa, Ma, or Ea determining the ovarall performance of alternative. We can simply say, using experts' judgments, we designed a Petri net that combines the scored PFoMs and gives the overall performance of the alternatives.

Fig. 8. Petri net model for determining overall performance of IT Infrastructure; P, M, and E represent Poor, Moderate, and Excellent, respectively; the index of a stands for Alternative performance of system. The same procedure stays for determining the overall cost of each alternative based on its scored CFoMs (EPC and O&M); except for overall cost, since the effects of EPC and O&M are significantly different in high values for different possession and payment policies, the output values for overall cost of each alternative is determined by two letters which correspond to EPC and O&M, respectively. Note that quantization levels for CFoMs are the same as that of performance FoMs, that is, Unusable, Moderate, and Excellent. Therefore, the overall cost of each alternative is U, MM, ME, EM, or EE (for Unusable values of CFoMs, we used only one output). Fig. 9 shows the designed Petri net for determining the overall cost of alternatives. Fig. 9. Petri net model for determining the cost of IT Infrastructure; U, M, and E represent Unusable, Moderate, and Excellent, respectively.

Now that we have the overall performance and cost characteristics of suggested IT infrastructures, we can compare each of them with the characteristics of the given policies. To do so, first, we consider the performance of the alternatives. Fig. 10 shows the proposed Petri net model that determines how much each alternative satisfies the wanted performance in given policy. As shown, based on the cost performance trend, three levels are considered for wanted performance, namely, Not Considered (NC) for cost-oriented policy, Moderate (M) for cost-performance equilibrium policy, and Excellent (E) for performance-oriented policy. In modeling method, one place is considered for each of these values; thus, based on what value is wanted in the given policy, the corresponding place gets one token and the other two places remain empty in initial marking. Outputs of the net of Fig. 9 (Pa, Ma, and Ea) are also inputs of this net. Using the rule that the suggested IT infrastructure should satisfy the wanted performance given by policy, we designed the Petri net of Fig. 9 for determining the level of satisfaction of given policy regarding performance characteristics by each IT infrastructure (for example, an IT infrastructure with Excellent performance satisfies all levels of wanted performance in an Excellent way). As shown, three levels of satisfaction are considered, Poor (P), Moderate (M), and Excellent (E), each one modeled by a place in the net. Fig. 10. Petri net model for applying policy regarding performance; NC, M, and E represent Not Considered, Moderate, and Excellent, respectively; the index of p stands for Policy. Fig. 11 shows the designed Petri net that determines the level of adaptation of each alternative with the type of possession given in policy. Left-hand side places are the outputs of the net of Fig. 10 showing the cost characteristics of the alternative, and upper places are considered for the possession policy, i.e. FP, MP, and LP correspond to full possession, moderate possession, and low possession, respectively (possession policy determines which

II. TECHNICAL OPPORTUNITIES AND CHALLENGES

The projected application of large-scale smart grid may be monumental. Various types of data can be collected to support generation, transmission, and distribution system modeling efforts. Collected data could be tagged with metadata and annotations to foster data sharing. Metadata consist of additional information that relates, for example, to the ownership, interpretation, and use of the data. Such information could help data use, security, and validation. Annotations to data can be made to assist in data searching and characterization to facilitate collaboration and workflow. Such collaborative data tagging is becoming increasingly common in science and engineering collaboration environments and provides a baseline for future energy system applications. A broad range of data that could potentially be collected includes (1) network topology and system component connectivity, (2) equipment parameters and operating status, and (3) anomalous network conditions caused by either scheduled or nonscheduled events. Once data is successfully collected, an additional set of data access concerns includes (1) data management, (2) dispute resolution of data sets, (3) data access challenges, (4) regard to proprietary data protection, (5) data integration quality challenges, and (6) data security and protection. In a similar manner, the distribution of simulation results and reports that are derived from these data sources needs to be carefully controlled, for they could convey sensitive information. This aspect of derived data management focuses on these sensitivities by identifying additional access, integrity, validation and version control, and security and protection concerns. In a large-scale application, data could be collected to characterize the status of available intermittent power sources such as wind and solar, plug-in hybrid electric vehicles, and new demand patterns resulting from consumers using smart grid technologies to make purchasing and conservation decisions. These diverse applications expand the number of data collection points well beyond the current number monitored in transmission and distribution grids. As a result, centrally or distributed data management practices potentially become a substantial task that requires further examination and research. As applications in system planning and operations incorporate these new system parameters, tools in the operations environment could take on more automated response as operators become comfortable with new automated capabilities. Because modeling algorithms and results can be improved when additional or better data are available, the motivation to implement smart grid data collection is elevated. However, barriers often exist that limit the availability, suitability, and unrestricted use of data. Although smart grid can simplify the gathering of such data, many other obstacles are still common and remain a focus of discussion. A key component required to resolve data integration and validation concerns may be handled by improved state estimation techniques. The challenges to the application of state estimation include more data points, more frequent data samples, and more potential data conflicts introduced both by multiple data sources and by an increase in the number of data points. Although ultimately mitigated using phase-shifting transformers, reactive power support equipment, new generation facilities, and transmission line enhancements or additions, additional data points and improved monitoring and simulation capabilities provide an opportunity to limit potential contingencies caused by loop flows while larger expansion projects are being assessed, proposed, and constructed. Improved tracking and situational awareness of network components provide fundamental data for more advanced analysis and visualization software to assist operator decision making. Such data improvements would also likely improve response to situations that could cause larger system consequences. For example, improved monitoring and tracking can expand data collection and foster the implementation of remedial action schemes (RASs). Improvements to RASs and other mitigation systems result in timely automatic response to further minimize undesirable system impacts caused by contingencies. Furthermore, traditional RAS applications generally operate on equipment located within a substation or a localized portion of a utility. RAS algorithms also often require manual equipment adjustments to improve performance and response. Because smart grid collects data across the entire network and provides access to data parameters and equipment controls, the potential for interregional RAS applications increases. Such an improvement in RAS scope is more likely to mitigate the negative effects caused by parallel flows, voltage operating violations, and contingencies that could cause large regional impacts. An additional benefit of large-scale monitoring is that system parameter modifications can be made at distant facilities that may best correct a problem with local impacts. Thus, the scope and comprehensiveness of RAS applications are expanded by smart grid. The coupling of improved data collection and advanced response applications further improves system mitigation measures and decreases adverse system responses like unnecessary load curtailments, voltage sensitivities, and other power quality concerns. For example, smart grid technologies can address transmission congestion issues through demand response and controllable load. However, along with considerable improvements in response capabilities comes an increase in computational and algorithmic challenges that identify additional development

opportunities. An influx of more data also has the potential to improve many aspects of system planning. For example, improved generation dispatch and line loading may be likely as dayahead forecasts are validated with current information from actual data points. In addition to the dispatch challenges of existing technologies, new challenges introduced by intermittent power sources, plug-in hybrid electric vehicles, 3 and dynamic customer demand patterns further increase the reliance on additional data to improve generation forecasts. An increase in the data collection of equipment parameters and operating status provides improved failure and maintenance information that can be used to refine equipment maintenance schedules. Equipment that is not in service affects the operation of other generators and transmission lines required to satisfy projected supply and demand schedules. Over committing generators has a large impact on system operating costs, so improved maintenance schedule estimation can have a significant impact on reducing utility costs. There are numerous software products to assist operators with monitoring and control tasks. Smart grid can expand the usefulness of these systems by increasing the number of represented local data points and by providing access to data that goes beyond the boundaries of the operating utility. While inclusion of more local data improves model results, inclusion of regional data supports the mitigation of regional contingencies that may have a significant local effect. These tools already assist operators; however, the capabilities of these tools may be potentially expanded. Increased data may also signify a challenge to software vendors as computational requirements increase as the number of data points and sampling rates increase.

III. INTELLIGENCE AND CONTROL OF INDIVIDUAL NODES

In this section we discuss concepts for creating the "Intelligence" blocks which define the autonomous behaviours of the distributed component. The (previously centralised) intelligence for coordinating all the components of the substation is now distributed across these components. Instead of simply passing of all information to the next level of hierarchy, each component makes a decision by itself as to whether the available information is sufficient, and informs higher level about the results. The decision is made based on the information available; if the accessible data is not satisfactory to make a decision then the information is passed to higher levels and authority to decide is given to them. This decentralization empowering the low levels simplifies the decision making algorithms while giving more independence to the components and makes the system more flexible and more easily reconfigurable without considerable

changes in the operating algorithms. At this stage of the research the following assumptions are made to simplify the collaborative algorithm.

1. A sectionalising switch can only be connected to one downstream and one upstream sectionalising switch.
2. A sectionalising switch can be connected to a single downstream tie switch.
3. A tie switch can only be connected to two upstream sectionalising switches.
4. An overcurrent relay can communicate with one downstream overcurrent relay.

Primary equipment does not perform complex behaviour; it sets initial position, responds to requests from bay level LNs, and makes simple decisions based on the available information, letting the upper layer know what decision has been made instead of transmitting data over the bus. The bay level LNs are distributed and need to interact with their neighbours to analyse the situation and make a decision. They require more "complex" intelligence. 6 As mentioned previously, there are sectionalising switches and tie switches, which differ in their purpose in the scheme and as a result in their behaviour algorithms. The important difference is that sectionalising switches are used to isolate faults, whereas tie switches are used to find an alternative source of supply on request. There are two layers in the bay level. The layer of PIOC LNs locates the fault. LNs within this layer "talk" to each other to determine the fault position, and provide this information to upper layer. The upper layer consists of CSWI LNs, which collaborate with each other, and supply tie switches with data necessary for alternative supply evaluation.

A. CSWI Intelligence (sectionalising switch)

CSWI has two modes of operation: normal state and fault state. When the section where a switch is located does not have fault the switch is operated in the normal state. This applies even if there is a fault in another part of distribution network; however the switch moves to the fault state if it is involved in the alternative supply restoration process (Figure 7). When the feeder that the switch belongs to has a fault, then the switch moves to the fault state. Figure 7 demonstrates the concept. Initially the CSWI is in the normal state. When PIOC replicates the LOCKOUT signal received from RREC to the connected CSWIs, those switches move to fault state; also when the tie switch has been commanded to restore supply. When the fault has been repaired, the substation is commanded to return to the pre-fault state.

Figure 7. Algorithm defining CSWI intelligence

In normal mode a sectionalising switch only collaborates with its upstream neighbour and the

downstream tie switch. By request of the tie switch, the upstream sectionalizing switches propagate a headroom request signal and pass down the calculated headroom value (calculated according to the method given in [1]).

In fault mode a sectionalising switch only talks to its downstream neighbour and the tie switch. In this mode any action and events related to headroom calculation are ignored. The switch which has a fault on its section of the feeder will isolate the fault by opening the adjacent downstream switch and controlled switch, and inform the adjacent downstream switch that the fault is isolated. The switch that does not have a fault, after the fault has been isolated, will initiate a search for and restore from an alternative source of supply.

B. CSWI Intelligence (tie switch)

A tie switch collaborates with both upstream sectionalizing switches. One of the sectionalising switches sends a request for alternative supply and the tie switch "negotiates" about supply restoration. The other sectionalising switch replies to enquiries about excess capacity. Based on this data the tie switch decides whether or not to "offer" supply to the requesting sectionalising switch.

C. PIOC Intelligence

PIOC detects and locates the fault, provides related information to the corresponding CSWI, and propagates the LOCKOUT signal. It triggers PTRC.Op.general data if there is a fault on the feeder. If there is a permanent fault, RREC goes to lockout and sends the LOCKOUT signal to PIOC, which replicates the LOCKOUT signal to let the downstream switches know about permanent fault somewhere on the feeder and initiate fault location algorithm. It senses the current with defined frequency and applies predefined rules to detect the fault. If monitored current was within acceptable limits before supply was interrupted then there is no fault on its section of the feeder. If a fault is detected it provides this status information. It keeps the pre-fault value of the current. It collaborates with the downstream PIOC, requesting fault status in order to locate the fault. Based on the data obtained it decides whether the fault is on its section or the section below.

D. TCTR Intelligence

The purpose of TCTR is to sample the current and provide the samples to PIOC.

E. PTRC Intelligence

PTRC sees that the Op.general has been triggered and issues a trip signal (Tr.general) to the corresponding switch controller.

F. RREC Intelligence

The OpOpn.general input of RREC is triggered by CSWI in case of a fault. This makes RREC move to "fault" state, where it performs preconfigured behaviour. The behaviour is simply a timer; when it expires RREC tries to reclose XCBR. If the attempt fails, RREC goes to the lockout state. It is restored to normal state by the "restore pre-fault state" command.

IV. MODELING FRAMEWORK

Traditionally, models created for optimization of systems are generally expressed as abstract mathematical models. These models are defined in standard mathematical lexicon. When a system is to be deployed, its model is realized as code segments or equation matrices or equation arrays based on solver being used for optimization. The dimensions of these matrices and cardinality of variables is usually defined at the time of deployment and is hard coded in code segments, matrix dimensions, etc. In comparison, for systems such as DAS, system dimensions at the time of deployment are meaningless . This is due to the fact that it can grow, as well as shrink over time. To handle such changes, a measure of self-aware modeling integrated with self-optimization is necessary to manage DAS. This selfaware optimization can leverage the change in dimensions of DAS at runtime to attain scalability and performance boost according to the runtime state of DAS. Various systems have been optimized through mathematical models. However, in all of the applications of mathematical techniques seen so far by the authors, the constraints and tuning parameters were known when the system was being implemented [5]. We have not observed any detailed work for engineering a system's model that exhibited variability in the size of their constraints and control features. Therefore in our modeling framework we have used the abstract mathematical models as a meta-model to create an on-demand, instantaneous model of system based on system statistics. In this section we define our modeling framework for constructing an instantaneous model of a system at runtime.

A. Structure of the Mathematical Meta-Model

In practice mathematical models are developed and expressed as abstract models. Mathematical models represent a 1564 system in form of decision variables and constraints. Decision variables are the controlling parameters to change the system state where as constraints are the limitations of the system. Since in mathematics, a variable can take any numeric value, it is important that we specify the limits of our decision variables as well. To model a system, the control parameters and limitations of the system are analyzed. A system can be composed of many control parameters but usually

there exist logical groupings with which these control parameters can be abstracted into a single entity or class. Usually this also means that similar constraints apply on each of the element of the grouping. It also means that a single abstract equation with appropriate quantifiers can suffice for containing the behavior of all the variables within a group. Since these are logical groupings and resemble a set like structure, we call these variable abstraction as ontologies of our system. Hence an ontology is a group of control parameters which have similar logical structure and are subjected to similar constraints. Like sets, ontologies can be grouped together to form more inclusive notation. Mathematically, this means that whereas two different logical groups of variables, or ontologies are subjected to their own constraints, it can also have a set of constraints that are applicable to both the groups. Hence our decision variables can be part of a multitude of ontologies. Here a subscript define the specific element within an ontology. We call these grouping of ontologies as an ontological class. Figure 1 describes the abstract model that we will discuss in detail here.

Example: Model for a Smart Grid Application
We take the example of modeling the usage of electrical devices in an electric grid. We divide our devices into ontologies according to their consumption profiles and time periods. Our task is to maximize the number of machines from each set which can be kept in "on" state for a particular period in an hour without violating the service-level guarantee. Here the number of machines to keep in "on" state in a particular time period is our tuning parameter or "decision variable". For each tuning parameter there are two ontologies. First there are different sets of machines. Each type is represented as a subscript i. The second attribute is of time, that is which time period does a specific decision variable represent. These types are represented as a subscript t. Hence i, and t represent two ontologies combined in a single decision variable $X_{i,t}$. The system in figure 1 is subjected to three classes of constraints. Each of these class is represented as a single abstract equation. Notice that equation 3 is only applicable to one ontology, the time t while the other two are subjected to both. For demonstration of our framework we will consider the example of equation 2 in detail. This equation constraints the system by enforcing a minimum service level. It states that for every time period t, the number of machines switched on in every machine class i should not be less than I_{j}3rd of the total number of machines in that class. During implementation these abstract models are expanded according to available system statistics. If our system had fixed machine classes, say 10 and 6 time periods (t) the abstract equation 2 would have been expanded to 60

equations. Each of these 60 equations would have represented one specific (t, i) tuple.

Mathematical models for systems which do not exhibit change in cardinality from abstract model to implemented model can be modeled effectively. That is, if we can enumerate at time of implementation or deployment as to how many machines we have and how many time segments we have, then generating an actual model of the system from abstract model is straight forward. However, if the cardinality cannot be evaluated at the time of implementation, then modeling becomes a difficult task. A naive modeling technique is to consider worst case scenario. For example, in the sample model above, we limit i, or device classes, to say 1000 and then make a model for these many classes. For a grid level electric distribution network this solution is not feasible. First, the number of device classes cannot be predicted. There are new types of machines that are being added everyday and limiting this growth is not possible. Second, worst case setup is highly inefficient. By calculating for a 1000 classes always, we are consuming much more resources where as in actuality we might need a fraction of these calculations. Third, because we always assume a large data-set, the choices for algorithms is limited. There are algorithms which are more efficient for small to medium sized data-sets. If we can evaluate and model at runtime, it is possible to derive a better result by using more accurate algorithms.

B. Modeling at Runtime
Various techniques exist for creating a runtime
We consider the following anycast field equations defined over an open bounded piece of network and /or feature space $\Omega \subset R^d$. They describe the dynamics of the mean anycast of each of p node populations.

$$\begin{cases} (\dfrac{d}{dt}+l_i)V_i(t,r) = \sum_{j=1}^{p}\int_{\Omega} J_{ij}(r,\bar{r})S[(V_j(t-\tau_{ij}(r,\bar{r}),\bar{r})-h_{|j})]d\bar{r} \\ \qquad\qquad +I_i^{ext}(r,t), \qquad t\geq 0, 1\leq i \leq p, \\ V_i(t,r)=\phi_i(t,r) \qquad\qquad t\in[-T,0] \end{cases} \quad (1)$$

We give an interpretation of the various parameters and functions that appear in (1), Ω is finite piece of nodes and/or feature space and is represented as an open bounded set of R^d. The vector r and $\bar{r}$ represent points in Ω. The function $S: R \rightarrow (0,1)$ is the normalized sigmoid function:

$$S(z) = \frac{1}{1+e^{-z}} \qquad (2)$$

It describes the relation between the input rate v_i of population i as a function of the packets potential, for example, $V_i = v_i = S[\sigma_i(V_i - h_i)]$. We note V the $p-$ dimensional vector $(V_1,...,V_p)$. The p function $\phi_i, i = 1,...,p,$ represent the initial conditions, see below. We note ϕ the $p-$ dimensional vector $(\phi_1,...,\phi_p)$. The p function $I_i^{ext}, i = 1,...,p,$ represent external factors from other network areas. We note I^{ext} the $p-$ dimensional vector $(I_1^{ext},...,I_p^{ext})$. The $p \times p$ matrix of functions $J = \{J_{ij}\}_{i,j=1,...,p}$ represents the connectivity between populations i and j, see below. The p real values $h_i, i = 1,...,p,$ determine the threshold of activity for each population, that is, the value of the nodes potential corresponding to 50% of the maximal activity. The p real positive values $\sigma_i, i = 1,...,p,$ determine the slopes of the sigmoids at the origin. Finally the p real positive values $l_i, i = 1,...,p,$ determine the speed at which each anycast node potential decreases exponentially toward its real value. We also introduce the function $S : R^p \to R^p$, defined by $S(x) = [S(\sigma_1(x_1 - h_1)),...,S(\sigma_p - h_p))]$, and the diagonal $p \times p$ matrix $L_0 = diag(l_1,...,l_p)$. Is the intrinsic dynamics of the population given by the linear response of data transfer. $(\frac{d}{dt} + l_i)$ is replaced by $(\frac{d}{dt} + l_i)^2$ to use the alpha function response. We use $(\frac{d}{dt} + l_i)$ for simplicity although our analysis applies to more general intrinsic dynamics. For the sake, of generality, the propagation delays are not assumed to be identical for all populations, hence they are described by a matrix $\tau(r, \bar{r})$ whose element $\tau_{ij}(r, \bar{r})$ is the propagation delay between population j at $\bar{r}$ and population i at r. The reason for this assumption is that it is still unclear from anycast if propagation delays are independent of the populations. We assume for technical reasons that τ is continuous, that is $\tau \in C^0(\bar{\Omega}^2, R_+^{p \times p})$. Moreover packet data indicate that τ is not a symmetric function i.e., $\tau_{ij}(r, \bar{r}) \neq \tau_{ij}(\bar{r}, r)$, thus no assumption is made about this symmetry unless otherwise stated. In order to compute the righthand side of (1), we need to know the node potential factor V on interval $[-T, 0]$. The value of T is obtained by considering the maximal delay:

$$\tau_m = \max_{i,j(r, \bar{r} \in \Omega \times \Omega)} \tau_{i,j}(r, \bar{r}) \qquad (3)$$

Hence we choose $T = \tau_m$

C. *Mathematical Framework*
A convenient functional setting for the non-delayed packet field equations is to use the space $F = L^2(\Omega, R^p)$ which is a Hilbert space endowed with the usual inner product:

$$\langle V, U \rangle_F = \sum_{i=1}^{p} \int_{\Omega} V_i(r)U_i(r)dr \qquad (1)$$

To give a meaning to (1), we defined the history space $C = C^0([-\tau_m, 0], F)$ with $\|\phi\| = \sup_{t \in [-\tau_m, 0]} \|\phi(t)\| F$, which is the Banach phase space associated with equation (3). Using the notation $V_t(\theta) = V(t + \theta), \theta \in [-\tau_m, 0]$, we write (1) as

$$\begin{cases} \dot{V}(t) = -L_0 V(t) + L_1 S(V_t) + I^{ext}(t), \\ \qquad V_0 = \phi \in C, \end{cases} \qquad (2)$$

Where

$$\begin{cases} L_1 : C \to F, \\ \phi \to \int_{\Omega} J(., \bar{r})\phi(\bar{r}, -\tau(., \bar{r}))d\bar{r} \end{cases}$$

Is the linear continuous operator satisfying $\|L_1\| \leq \|J\|_{L^2(\Omega^2, R^{p \times p})}$. Notice that most of the papers on this subject assume Ω infinite, hence requiring $\tau_m = \infty$.

Proposition 1.0 If the following assumptions are satisfied.

1. $J \in L^2(\Omega^2, R^{p \times p})$,

2.   The external current $I^{ext} \in C^0(R, F)$,

3. $\tau \in C^0(\bar{\Omega}^2, R_+^{p \times p}), \sup_{\bar{\Omega}^2} \tau \leq \tau_m$.

Then for any $\phi \in C$, there exists a unique solution $V \in C^1([0, \infty), F) \cap C^0([-\tau_m, \infty, F)$ to (3)

Notice that this result gives existence on R_+, finite-time explosion is impossible for this delayed differential equation. Nevertheless, a particular solution could grow indefinitely, we now prove that this cannot happen.

D. *Boundedness of Solutions*

A valid model of neural networks should only feature bounded packet node potentials.

Theorem 1.0 All the trajectories are ultimately bounded by the same constant R if

$$I \equiv \max_{t \in R^+} \left\| I^{ext}(t) \right\|_F < \infty.$$

Proof : Let us defined $f : R \times C \to R^+$ as

$$f(t, V_t) \stackrel{def}{=} \left\langle -L_0 V_t(0) + L_1 S(V_t) + I^{ext}(t), V(t) \right\rangle_F = \frac{1}{2} \frac{d \|V\|_F^2}{dt}$$

We note $l = \min_{i=1,\dots p} l_i$

$$f(t, V_t) \leq -l \|V(t)\|_F^2 + (\sqrt{p|\Omega|} \|J\|_F + I) \|V(t)\|_F$$

Thus, if

$$\|V(t)\|_F \geq 2 \frac{\sqrt{p|\Omega|}.\|J\|_F + I}{l} \stackrel{def}{=} R, f(t, V_t) \leq -\frac{lR^2}{2} \stackrel{def}{=} -\delta < 0$$

Let us show that the open route of F of center 0 and radius R, B_R, is stable under the dynamics of equation. We know that $V(t)$ is defined for all $t \geq 0s$ and that $f < 0$ on ∂B_R, the boundary of B_R. We consider three cases for the initial condition V_0. If $\|V_0\|_C < R$ and set $T = \sup\{t \mid \forall s \in [0,t], V(s) \in \overline{B_R}\}$. Suppose that $T \in R$, then $V(T)$ is defined and belongs to $\overline{B_R}$, the closure of B_R, because $\overline{B_R}$ is closed, in effect to ∂B_R, we also have

$$\frac{d}{dt} \|V\|_F^2 \mid_{t=T} = f(T, V_T) \leq -\delta < 0 \qquad \text{because}$$

$V(T) \in \partial B_R$. Thus we deduce that for $\varepsilon > 0$ and small enough, $V(T + \varepsilon) \in \overline{B_R}$ which contradicts the definition of T. Thus $T \notin R$ and $\overline{B_R}$ is stable.

Because f<0 on $\partial B_R, V(0) \in \partial B_R$ implies that $\forall t > 0, V(t) \in B_R$. Finally we consider the case $V(0) \in C\overline{B_R}$. Suppose that $\forall t > 0, V(t) \notin \overline{B_R}$, then

$$\forall t > 0, \frac{d}{dt} \|V\|_F^2 \leq -2\delta, \qquad \text{thus} \qquad \|V(t)\|_F \quad \text{is}$$

monotonically decreasing and reaches the value of R in finite time when $V(t)$ reaches ∂B_R. This contradicts our assumption. Thus $\exists T > 0 \mid V(T) \in B_R$.

Proposition 1.1 : Let s and t be measured simple functions on X. for $E\varepsilon M$, define

$$\phi(E) = \int_E s \, d\mu \qquad (1)$$

Then ϕ is a measure on M.

$$\int_X (s+t) d\mu = \int_X s \, d\mu + \int_X t \, d\mu \qquad (2)$$

Proof : If s and if $E_1, E_2, \dots$ are disjoint members of M whose union is E, the countable additivity of μ shows that

$$\phi(E) = \sum_{i=1}^n \alpha_i \mu(A_i \cap E) = \sum_{i=1}^n \alpha_i \sum_{r=1}^\infty \mu(A_i \cap E_r)$$

$$= \sum_{r=1}^\infty \sum_{i=1}^n \alpha_i \mu(A_i \cap E_r) = \sum_{r=1}^\infty \phi(E_r)$$

Also, $\varphi(\phi) = 0,$ so that φ is not identically ∞.

Next, let s be as before, let $\beta_1, \dots, \beta_m$ be the distinct values of t, and let $B_j = \{x : t(x) = \beta_j\}$ If $E_{ij} = A_i \cap B_j$, the

$$\int_{E_{ij}} (s+t) d\mu = (\alpha_i + \beta_j) \mu(E_{ij})$$

and $\int_{E_{ij}} s \, d\mu + \int_{E_{ij}} t \, d\mu = \alpha_i \mu(E_{ij}) + \beta_j \mu(E_{ij})$

Thus (2) holds with E_{ij} in place of X. Since X is the disjoint union of the sets E_{ij} $(1 \leq i \leq n, 1 \leq j \leq m)$, the first half of our proposition implies that (2) holds.

Theorem 1.1: If K is a compact set in the plane whose complement is connected, if f is a continuous complex function on K which is holomorphic in the interior of , and if $\varepsilon > 0$, then there exists a polynomial P such that $|f(z) = P(z)| < \varepsilon$ for all $z\varepsilon K$. If the interior of K is empty, then part of the hypothesis is vacuously satisfied, and the conclusion holds for every $f \varepsilon C(K)$. Note that K need to be connected.

Proof: By Tietze's theorem, f can be extended to a continuous function in the plane, with compact support. We fix one such extension and denote it again by f. For any $\delta > 0$, let $\omega(\delta)$ be the supremum of the numbers $|f(z_2) - f(z_1)|$ Where

z_1 and z_2 are subject to the condition $|z_2 - z_1| \le \delta$. Since f is uniformly continous, we have $\lim_{\delta \to 0} \omega(\delta) = 0$ $\qquad$ (1) From now on, δ will be fixed. We shall prove that there is a polynomial P such that

$$|f(z) - P(z)| < 10,000 \, \omega(\delta) \quad (z \varepsilon K) \qquad (2)$$

By (1), this proves the theorem. Our first objective is the construction of a function $\Phi \varepsilon C_c'(R^2)$, such that for all z

$$|f(z) - \Phi(z)| \le \omega(\delta), \qquad (3)$$

$$|(\partial \Phi)(z)| < \frac{2\omega(\delta)}{\delta}, \qquad (4)$$

And

$$\Phi(z) = -\frac{1}{\pi} \iint_X \frac{(\partial \Phi)(\zeta)}{\zeta - z} d\zeta d\eta \qquad (\zeta = \xi + i\eta), \qquad (5)$$

Where X is the set of all points in the support of Φ whose distance from the complement of K does not δ. (Thus X contains no point which is "far within" K.) We construct Φ as the convolution of f with a smoothing function A. Put $a(r) = 0$ if $r > \delta$, put

$$a(r) = \frac{3}{\pi \delta^2}(1 - \frac{r^2}{\delta^2})^2 \qquad (0 \le r \le \delta), \qquad (6)$$

And define

$$A(z) = a(|z|) \qquad (7)$$

For all complex z. It is clear that $A \varepsilon C_c'(R^2)$. We claim that

$$\iint_{R^s} A = 1, \qquad (8)$$

$$\iint_{R^2} \partial A = 0, \qquad (9)$$

$$\iint_{R^3} |\partial A| = \frac{24}{15\delta} < \frac{2}{\delta}, \qquad (10)$$

The constants are so adjusted in (6) that (8) holds. (Compute the integral in polar coordinates), (9) holds simply because A has compact support. To compute (10), express ∂A in polar coordinates, and note that $\frac{\partial A}{\partial \theta} = 0$,

$$\frac{\partial A}{\partial r} = -a',$$

Now define

$$\Phi(z) = \iint_{R^2} f(z - \zeta) A d\xi d\eta = \iint_{R^2} A(z - \zeta) f(\zeta) d\xi d\eta \qquad (11)$$

Since f and A have compact support, so does Φ. Since

$$\Phi(z) - f(z)$$
$$= \iint_{R^2} [f(z - \zeta) - f(z)] A(\xi) d\xi d\eta \quad (12)$$

And $A(\zeta) = 0$ if $|\zeta| > \delta$, (3) follows from (8). The difference quotients of A converge boundedly to the corresponding partial derivatives, since $A \varepsilon C_c'(R^2)$. Hence the last expression in (11) may be differentiated under the integral sign, and we obtain

$$(\partial \Phi)(z) = \iint_{R^2} (\overline{\partial A})(z - \zeta) f(\zeta) d\xi d\eta$$
$$= \iint_{R^2} f(z - \zeta)(\partial A)(\zeta) d\xi d\eta$$
$$= \iint_{R^2} [f(z - \zeta) - f(z)](\partial A)(\zeta) d\xi d\eta \qquad (13)$$

The last equality depends on (9). Now (10) and (13) give (4). If we write (13) with Φ_x and Φ_y in place of $\partial \Phi$, we see that Φ has continuous partial derivatives, if we can show that $\partial \Phi = 0$ in G, where G is the set of all $z \varepsilon K$ whose distance from the complement of K exceeds δ. We shall do this by showing that

$$\Phi(z) = f(z) \qquad (z \varepsilon G); \qquad (14)$$

Note that $\partial f = 0$ in G, since f is holomorphic there. Now if $z \varepsilon G$, then $z - \zeta$ is in the interior of K for all ζ with $|\zeta| < \delta$. The mean value property for harmonic functions therefore gives, by the first equation in (11),

$$\Phi(z) = \int_0^\delta a(r) r dr \int_0^{2\pi} f(z - re^{i\theta}) d\theta$$
$$= 2\pi f(z) \int_0^\delta a(r) r dr = f(z) \iint_{R^2} A = f(z) \qquad (15)$$

For all $z \varepsilon G$, we have now proved (3), (4), and (5) The definition of X shows that X is compact and that X can be covered by finitely many open discs $D_1, ..., D_n$, of radius 2δ, whose centers are not in K. Since $S^2 - K$ is connected, the center of each D_j can be joined to ∞ by a polygonal path in $S^2 - K$. It follows that each D_j contains a compact

connected set E_j, of diameter at least 2δ, so that $S^2 - E_j$ is connected and so that $K \cap E_j = \phi$. with $r = 2\delta$. There are functions $g_j \varepsilon H(S^2 - E_j)$ and constants b_j so that the inequalities.

$$\left| Q_j(\zeta, z) \right| < \frac{50}{\delta}, \qquad (16)$$

$$\left| Q_j(\zeta, z) - \frac{1}{z - \zeta} \right| < \frac{4,000\delta^2}{\left| z - \zeta \right|^2} \qquad (17)$$

Hold for $z \notin E_j$ and $\zeta \in D_j$, if

$$Q_j(\zeta, z) = g_j(z) + (\zeta - b_j)g_j^2(z) \qquad (18)$$

Let Ω be the complement of $E_1 \cup ... \cup E_n$. Then Ω is an open set which contains K. Put $X_1 = X \cap D_1$ and $X_j = (X \cap D_j) - (X_1 \cup ... \cup X_{j-1})$, for $2 \le j \le n$,

Define

$$R(\zeta, z) = Q_j(\zeta, z) \qquad (\zeta \varepsilon X_j, z \varepsilon \Omega) \qquad (19)$$

And

$$F(z) = \frac{1}{\pi} \iint_X (\partial \Phi)(\zeta) R(\zeta, z) d\zeta d\eta \qquad (20)$$

$$(z \ \varepsilon \ \Omega)$$

Since,

$$F(z) = \sum_{j=1} \frac{1}{\pi} \iint_{X_i} (\partial \Phi)(\zeta) Q_j(\zeta, z) d\xi d\eta, \qquad (21)$$

(18) shows that F is a finite linear combination of the functions g_j and g_j^2. Hence $F \varepsilon H(\Omega)$. By (20), (4), and (5) we have

$$\left| F(z) - \Phi(z) \right| < \frac{2\omega(\delta)}{\pi\delta} \iint_X | R(\zeta, z)$$

$$- \frac{1}{z - \zeta} | d\xi d\eta \quad (z \ \varepsilon \ \Omega) \ (22)$$

Observe that the inequalities (16) and (17) are valid with R in place of Q_j if $\zeta \ \varepsilon \ X$ and $z \ \varepsilon \ \Omega$.

Now fix $z \ \varepsilon \ \Omega$, put $\zeta = z + \rho e^{i\theta}$, and estimate the integrand in (22) by (16) if $\rho < 4\delta$, by (17) if $4\delta \le \rho$. The integral in (22) is then seen to be less than the sum of

$$2\pi \int_0^{4\delta} \left(\frac{50}{\delta} + \frac{1}{\rho} \right) \rho d\rho = 808\pi\delta \qquad (23)$$

And

$$2\pi \int_{4\delta}^\infty \frac{4,000\delta^2}{\rho^2} \rho d\rho = 2,000\pi\delta. \qquad (24)$$

Hence (22) yields

$$\left| F(z) - \Phi(z) \right| < 6,000\omega(\delta) \qquad (z \ \varepsilon \ \Omega) \quad (25)$$

Since $F \ \varepsilon \ H(\Omega), K \subset \Omega,$ and $S^2 - K$ is connected, Runge's theorem shows that F can be uniformly approximated on K by polynomials. Hence (3) and (25) show that (2) can be satisfied. This completes the proof.

Lemma 1.0 : Suppose $f \varepsilon C_c^{'}(R^2)$, the space of all continuously differentiable functions in the plane, with compact support. Put

$$\partial = \frac{1}{2} \left(\frac{\partial}{\partial x} + i \frac{\partial}{\partial y} \right) \qquad (1)$$

Then the following "Cauchy formula" holds:

$$f(z) = -\frac{1}{\pi} \iint_{R^2} \frac{(\partial f)(\zeta)}{\zeta - z} d\xi d\eta$$

$$(\zeta = \xi + i\eta) \qquad (2)$$

Proof: This may be deduced from Green's theorem. However, here is a simple direct proof:

Put $\varphi(r, \theta) = f(z + re^{i\theta}), r > 0, \theta$ real

If $\zeta = z + re^{i\theta}$, the chain rule gives

$$(\partial f)(\zeta) = \frac{1}{2} e^{i\theta} \left[\frac{\partial}{\partial r} + \frac{i}{r} \frac{\partial}{\partial \theta} \right] \varphi(r, \theta) \qquad (3)$$

The right side of (2) is therefore equal to the limit, as $\varepsilon \to 0,$ of

$$-\frac{1}{2} \int_\varepsilon^\infty \int_0^{2\pi} \left(\frac{\partial \varphi}{\partial r} + \frac{i}{r} \frac{\partial \varphi}{\partial \theta} \right) d\theta dr \qquad (4)$$

For each $r > 0, \varphi$ is periodic in θ, with period 2π. The integral of $\partial \varphi / \partial \theta$ is therefore 0, and (4) becomes

$$-\frac{1}{2\pi} \int_0^{2\pi} d\theta \int_\varepsilon^\infty \frac{\partial \varphi}{\partial r} dr = \frac{1}{2\pi} \int_0^{2\pi} \varphi(\varepsilon, \theta) d\theta \qquad (5)$$

As $\varepsilon \to 0, \varphi(\varepsilon, \theta) \to f(z)$ uniformly. This gives (2)

If $X^\alpha \in a$ and $X^\beta \in k[X_1,...X_n]$, then $X^\alpha X^\beta = X^{\alpha+\beta} \in a$, and so A satisfies the condition $(*)$. Conversely,

$$(\sum_{\alpha \in A} c_\alpha X^\alpha)(\sum_{\beta \in \square^n} d_\beta X^\beta) = \sum_{\alpha,\beta} c_\alpha d_\beta X^{\alpha+\beta} \qquad (finite\ sums),$$

and so if A satisfies $(*)$, then the subspace generated by the monomials $X^\alpha, \alpha \in a$, is an ideal. The proposition gives a classification of the monomial ideals in $k[X_1,...X_n]$: they are in one to one correspondence with the subsets A of $\square^n$ satisfying $(*)$. For example, the monomial ideals in $k[X]$ are exactly the ideals (X^n), $n \ge 1$, and the zero ideal (corresponding to the empty set A). We write $\langle X^\alpha \mid \alpha \in A \rangle$ for the ideal corresponding to A (subspace generated by the $X^\alpha, \alpha \in a$).

LEMMA 1.1. Let S be a subset of $\square^n$. The the ideal a generated by $X^\alpha, \alpha \in S$ is the monomial ideal corresponding to

$$A \stackrel{df}{=} \{\beta \in \square^n \mid \beta - \alpha \in \square^n, \quad some\ \alpha \in S\}$$

Thus, a monomial is in a if and only if it is divisible by one of the $X^\alpha, \alpha \in \mid S$

PROOF. Clearly A satisfies $(*)$, and $a \subset \langle X^\beta \mid \beta \in A \rangle$. Conversely, if $\beta \in A$, then $\beta - \alpha \in \square^n$ for some $\alpha \in S$, and $X^\beta = X^\alpha X^{\beta-\alpha} \in a$. The last statement follows from the fact that $X^\alpha \mid X^\beta \Leftrightarrow \beta - \alpha \in \square^n$. Let $A \subset \square^n$ satisfy $(*)$. From the geometry of A, it is clear that there is a finite set of elements $S = \{\alpha_1,...\alpha_s\}$ of A such that $A = \{\beta \in \square^n \mid \beta - \alpha_i \in \square^2,\ some\ \alpha_i \in S\}$ (The $\alpha_i's$ are the corners of A) Moreover,

$$a \stackrel{df}{=} \langle X^\alpha \mid \alpha \in A \rangle$$ is generated by the monomials $X^{\alpha_i}, \alpha_i \in S$.

DEFINITION 1.0. For a nonzero ideal a in $k[X_1,...,X_n]$, we let $(LT(a))$ be the ideal generated by

$$\{LT(f) \mid f \in a\}$$

LEMMA 1.2 Let a be a nonzero ideal in $k[X_1,...,X_n]$; then $(LT(a))$ is a monomial ideal, and it equals $(LT(g_1),...,LT(g_n))$ for some $g_1,...,g_n \in a$.

PROOF. Since $(LT(a))$ can also be described as the ideal generated by the leading monomials (rather than the leading terms) of elements of a.

THEOREM 1.2. Every *ideal* a in $k[X_1,...,X_n]$ is finitely generated; more precisely, $a = (g_1,...,g_s)$ where $g_1,...,g_s$ are any elements of a whose leading terms generate $LT(a)$

PROOF. Let $f \in a$. On applying the division algorithm, we find $f = a_1 g_1 +...+ a_s g_s + r, \qquad a_i, r \in k[X_1,...,X_n]$, where either $r = 0$ or no monomial occurring in it is divisible by any $LT(g_i)$. But $r = f - \sum a_i g_i \in a$, and therefore $LT(r) \in LT(a) = (LT(g_1),...,LT(g_s))$, implies that every monomial occurring in r is divisible by one in $LT(g_i)$. Thus $r = 0$, and $g \in (g_1,...,g_s)$.

DEFINITION 1.1. A finite subset $S = \{g_1,\mid...,g_s\}$ of an ideal a is a standard ($(Gr\ddot{o}bner)$ bases for a if $(LT(g_1),...,LT(g_s)) = LT(a)$. In other words, S is a standard basis if the leading term of every element of a is divisible by at least one of the leading terms of the g_i.

THEOREM 1.3 *The ring* $k[X_1,...,X_n]$ *is Noetherian i.e., every ideal is finitely generated.*

PROOF. For $n = 1$, $k[X]$ is a principal ideal domain, which means that every ideal is generated by single element. We shall prove the theorem by induction on n. Note that the obvious map $k[X_1,...X_{n-1}][X_n] \to k[X_1,...X_n]$ is an isomorphism – this simply says that every polynomial f in n variables $X_1,...X_n$ can be

expressed uniquely as a polynomial in X_n with coefficients in $k[X_1,...,X_n]$:

$$f(X_1,...X_n) = a_0(X_1,...X_{n-1})X_n^r + ... + a_r(X_1,...X_{n-1})$$

Thus the next lemma will complete the proof

LEMMA 1.3. If A is Noetherian, then so also is $A[X]$

PROOF.   For a polynomial

$$f(X) = a_0 X^r + a_1 X^{r-1} + ... + a_r, \quad a_i \in A, \quad a_0 \neq 0,$$

r is called the degree of f, and a_0 is its leading coefficient. We call 0 the leading coefficient of the polynomial 0.   Let a be an ideal in $A[X]$. The leading coefficients of the polynomials in a form an ideal a' in A, and since A is Noetherian, a' will be finitely generated. Let $g_1,...,g_m$ be elements of a whose leading coefficients generate a', and let r be the maximum degree of g_i. Now let $f \in a$, and suppose f has degree $s > r$, say, $f = aX^s + ...$ Then $a \in a'$, and so we can write

$$a = \sum b_i a_i, \qquad b_i \in A,$$

$a_i = $ leading coefficient of g_i
Now

$$f - \sum b_i g_i X^{s-r_i}, \quad r_i = \deg(g_i),$$ has degree $< \deg(f)$. By continuing in this way, we find that $f \equiv f_t \qquad \mod(g_1,...g_m)$ With f_t a polynomial of degree $t < r$. For each $d < r$, let a_d be the subset of A consisting of 0 and the leading coefficients of all polynomials in a of degree d; it is again an ideal in A. Let $g_{d,1},...,g_{d,m_d}$ be polynomials of degree d whose leading coefficients generate a_d. Then the same argument as above shows that any polynomial f_d in a of degree d can be written $f_d \equiv f_{d-1} \qquad \mod(g_{d,1},...g_{d,m_d})$ With f_{d-1} of degree $\leq d-1$. On applying this remark repeatedly we find that $f_t \in (g_{r-1,1},...g_{r-1,m_{r-1}},...g_{0,1},...g_{0,m_0})$ Hence

$$f_t \in (g_1,...g_m g_{r-1,1},...g_{r-1,m_{r-1}},...,g_{0,1},...,g_{0,m_0})$$

and so the polynomials $g_1,...,g_{0,m_0}$ generate a

One of the great successes of category theory in computer science has been the development of a "unified theory" of the constructions underlying denotational semantics. In the untyped λ-calculus, any term may appear in the function position of an application. This means that a model D of the λ-calculus must have the property that given a term t whose interpretation is $d \in D$, Also, the interpretation of a functional abstraction like $\lambda x . x$ is most conveniently defined as a function from $D\,to\,D$, which must then be regarded as an element of D. Let $\psi:[D \rightarrow D] \rightarrow D$ be the function that picks out elements of D to represent elements of $[D \rightarrow D]$ and $\phi:D \rightarrow [D \rightarrow D]$ be the function that maps elements of D to functions of D. Since $\psi(f)$ is intended to represent the function f as an element of D, it makes sense to require that $\phi(\psi(f)) = f$, that is, $\psi o \psi = id_{[D \rightarrow D]}$ Furthermore, we often want to view every element of D as representing some function from $D\,to\,D$ and require that elements representing the same function be equal – that is $\psi(\varphi(d)) = d$

or

$$\psi\,o\,\phi = id_D$$

The latter condition is called extensionality. These conditions together imply that $\phi\,and\,\psi$ are inverses--- that is, D is isomorphic to the space of functions from $D\,to\,D$ that can be the interpretations of functional abstractions: $D \cong [D \rightarrow D]$. Let us suppose we are working with the untyped $\lambda - calculus$, we need a solution ot the equation $D \cong A + [D \rightarrow D]$, where A is some predetermined domain containing interpretations for elements of C. Each element of D corresponds to either an element of A or an element of $[D \rightarrow D]$, with a tag. This equation can be solved by finding least fixed points of the function $F(X) = A + [X \rightarrow X]$ from domains to domains --- that is, finding domains X such that $X \cong A + [X \rightarrow X]$, and such that for any domain Y also satisfying this equation, there is an embedding of X to Y --- a pair of maps

$$X \quad \overset{f}{\underset{f^R}{\square}} \quad Y$$

Such that

$$f^R \, o \, f = id_X$$

$$f \, o \, f^R \subseteq id_Y$$

Where $f \subseteq g$ means that *f approximates g* in some ordering representing their information content. The key shift of perspective from the domain-theoretic to the more general category-theoretic approach lies in considering *F* not as a function on domains, but as a *functor* on a category of domains. Instead of a least fixed point of the function, *F*.

Definition 1.3: Let **K** be a category and $F : K \rightarrow K$ as a functor. A fixed point of *F* is a pair (A,a), where A is a **K-object** and $a : F(A) \rightarrow A$ is an isomorphism. A prefixed point of F is a pair (A,a), where A is a **K-object** and a is any arrow from F(A) to A

Definition 1.4 : An $\omega - chain$ in a category **K** is a diagram of the following form:

$$\Delta = D_o \xrightarrow{f_o} D_1 \xrightarrow{f_1} D_2 \xrightarrow{f_2} .....$$

Recall that a cocone μ of an $\omega - chain$ Δ is a *K*-object *X* and a collection of K –arrows $\{\mu_i : D_i \rightarrow X \mid i \geq 0\}$ such that $\mu_i = \mu_{i+1} \, o \, f_i$ for all $i \geq 0$. We sometimes write $\mu : \Delta \rightarrow X$ as a reminder of the arrangement of $\mu's$ components Similarly, a colimit $\mu : \Delta \rightarrow X$ is a cocone with the property that if $\nu : \Delta \rightarrow X'$ is also a cocone then there exists a unique mediating arrow $k : X \rightarrow X'$ such that for all $i \geq 0,, v_i = k \, o \, \mu_i$. Colimits of $\omega - chains$ are sometimes referred to as $\omega - co\lim its$. Dually, an $\omega^{op} - chain$ in *K* is a diagram of the following form:

$$\Delta = D_o \xleftarrow{f_o} D_1 \xleftarrow{f_1} D_2 \xleftarrow{f_2} .....$$

A cone $\mu : X \rightarrow \Delta$ of an $\omega^{op} - chain$ Δ is a **K**-object X and a collection of **K**-arrows $\{\mu_i : D_i \mid i \geq 0\}$ such that for all $i \geq 0$, $\mu_i = f_i \, o \, \mu_{i+1}$. An ω^{op} -limit of an $\omega^{op} - chain$ Δ is a cone $\mu : X \rightarrow \Delta$ with the property that if $\nu : X' \rightarrow \Delta$ is also a cone, then there exists a unique mediating arrow $k : X' \rightarrow X$ such that for all $i \geq 0, \mu_i \, o \, k = v_i$. We write $\perp_k$ (or just $\perp$) for the distinguish initial object of **K,** when it has one, and $\perp \rightarrow A$ for the unique arrow from $\perp$ to each **K**-object A. It is also convenient to write $\Delta^- = D_1 \xrightarrow{f_1} D_2 \xrightarrow{f_2} .....$ to denote all of Δ

except D_o and f_0. By analogy, μ^- is $\{\mu_i \mid i \geq 1\}$. For the images of Δ and μ under *F* we write

$$F(\Delta) = F(D_o) \xrightarrow{F(f_o)} F(D_1) \xrightarrow{F(f_1)} F(D_2) \xrightarrow{F(f_2)} .....$$

and $F(\mu) = \{F(\mu_i) \mid i \geq 0\}$

We write F^i for the *i*-fold iterated composition of *F* that is, $F^o(f) = f, F^1(f) = F(f), F^2(f) = F(F(f))$,etc. With these definitions we can state that every monitonic function on a complete lattice has a least fixed point:

Lemma 1.4. Let **K** be a category with initial object $\perp$ and let $F : K \rightarrow K$ be a functor. Define the $\omega - chain \Delta$ by

$$\Delta = \perp \xrightarrow{!\perp \rightarrow F(\perp)} F(\perp) \xrightarrow{F(!\perp \rightarrow F(\perp))} F^2(\perp) \xrightarrow{F^2(!\perp \rightarrow F(\perp))} ....$$

If both $\mu : \Delta \rightarrow D$ and $F(\mu) : F(\Delta) \rightarrow F(D)$ are colimits, then (D,d) is an intial F-algebra, where $d : F(D) \rightarrow D$ is the mediating arrow from $F(\mu)$ to the cocone μ^-

Theorem 1.4 Let a DAG G given in which each node is a random variable, and let a discrete conditional probability distribution of each node given values of its parents in G be specified. Then the product of these conditional distributions yields a joint probability distribution P of the variables, and (G,P) satisfies the Markov condition.

Proof. Order the nodes according to an ancestral ordering. Let $X_1, X_2, ...X_n$ be the resultant ordering. Next define.

$$P(x_1, x_2, ...x_n) = P(x_n \mid pa_n) P(x_{n-1} \mid Pa_{n-1})...$$
$$..P(x_2 \mid pa_2) P(x_1 \mid pa_1),$$

Where PA_i is the set of parents of X_i of in G and $P(x_i \mid pa_i)$ is the specified conditional probability distribution. First we show this does indeed yield a joint probability distribution. Clearly, $0 \leq P(x_1, x_2, ...x_n) \leq 1$ for all values of the variables. Therefore, to show we have a joint distribution, as the variables range through all their possible values, is equal to one. To that end, Specified conditional distributions are the conditional distributions they notationally represent in the joint distribution. Finally, we show the Markov condition is satisfied. To do this, we need show for $1 \leq k \leq n$ that

whenever

$$P(pa_k) \neq 0, if \ P(nd_k \mid pa_k) \neq 0$$

$$and \quad P(x_k \mid pa_k) \neq 0$$

then $P(x_k \mid nd_k, pa_k) = P(x_k \mid pa_k),$

Where ND_k is the set of nondescendents of X_k of in G. Since $PA_k \subseteq ND_k$, we need only show $P(x_k \mid nd_k) = P(x_k \mid pa_k)$. First for a given k , order the nodes so that all and only nondescendents of X_k precede X_k in the ordering. Note that this ordering depends on k , whereas the ordering in the first part of the proof does not. Clearly then

$$ND_k = \{X_1, X_2, \ldots X_{k-1}\}$$

$$Let$$

$$D_k = \{X_{k+1}, X_{k+2}, \ldots X_n\}$$

follows $\sum_{d_k}$

We define the mth *cyclotomic field to be the field* $Q[x]/(\Phi_m(x))$ *Where* $\Phi_m(x)$ *is the* m^{th} cyclotomic polynomial. $Q[x]/(\Phi_m(x))$ $\Phi_m(x)$ *has degree* $\varphi(m)$ *over* Q *since* $\Phi_m(x)$ has degree $\varphi(m)$. *The roots of* $\Phi_m(x)$ *are just the* primitive m^{th} roots of unity, so the complex embeddings of $Q[x]/(\Phi_m(x))$ *are simply the* $\varphi(m)$ *maps*

$$\sigma_k : Q[x]/(\Phi_m(x)) \mapsto C,$$

$$1 \le k \prec m, (k,m) = 1, \quad where$$

$$\sigma_k(x) = \xi_m^k,$$

ξ_m being our fixed choice of primitive m^{th} root of unity. Note that $\xi_m^k \in Q(\xi_m)$ for every k; it follows that $Q(\xi_m) = Q(\xi_m^k)$ for all k relatively prime to m . In particular, the images of the σ_i coincide, so $Q[x]/(\Phi_m(x))$ *is Galois over* Q . *This means that we can write* $Q(\xi_m)$ *for* $Q[x]/(\Phi_m(x))$ *without much fear of ambiguity; we will do so from now on, the identification being* $\xi_m \mapsto x.$ *One advantage of this is that one can easily talk about cyclotomic fields being extensions of one another, or intersections or compositums; all of these things take place considering them as subfield of* C. We now investigate some basic properties of cyclotomic fields. The first issue is whether or not they are all

distinct; to determine this, we need to know which roots of unity lie in $Q(\xi_m)$ .Note, for example, that if m is odd, then $-\xi_m$ is a $2m^{th}$ root of unity. We will show that this is the only way in which one can obtain any non-m^{th} roots of unity.

LEMMA 1.5 If m divides n , then $Q(\xi_m)$ *is contained in* $Q(\xi_n)$

PROOF. Since $\xi^{n/m} = \xi_m,$ *we have* $\xi_m \in Q(\xi_n),$ *so the result is clear*

LEMMA 1.6 If m *and* n *are relatively prime, then*

$$Q(\xi_m, \xi_n) = Q(\xi_{nm})$$

and

$$Q(\xi_m) \cap Q(\xi_n) = Q$$

(Recall the $Q(\xi_m, \xi_n)$ is the compositum of $Q(\xi_m)$ *and* $Q(\xi_n)$)

PROOF. One checks easily that $\xi_m \xi_n$ is a primitive mn^{th} root of unity, so that

$$Q(\xi_{mn}) \subseteq Q(\xi_m, \xi_n)$$

$$[Q(\xi_m, \xi_n) : Q] \le [Q(\xi_m) : Q][Q(\xi_n : Q]$$

$$= \varphi(m)\varphi(n) = \varphi(mn);$$

Since $[Q(\xi_{mn}) : Q] = \varphi(mn);$ this implies that $Q(\xi_m, \xi_n) = Q(\xi_{nm})$ We know that $Q(\xi_m, \xi_n)$ has degree $\varphi(mn)$ over Q , so we must have

$$[Q(\xi_m, \xi_n) : Q(\xi_m)] = \varphi(n)$$

and

$$[Q(\xi_m, \xi_n) : Q(\xi_m)] = \varphi(m)$$

$$[Q(\xi_m) : Q(\xi_m) \cap Q(\xi_n)] \ge \varphi(m)$$

And thus that $Q(\xi_m) \cap Q(\xi_n) = Q$

PROPOSITION 1.2 For any m and n

$$Q(\xi_m, \xi_n) = Q(\xi_{[m,n]})$$

And

$$Q(\xi_m) \cap Q(\xi_n) = Q(\xi_{(m,n)});$$

here $[m,n]$ and (m,n) denote the least common multiple and the greatest common divisor of m and n, respectively.

PROOF. Write $m = p_1^{e_1}......p_k^{e_k}$ and $p_1^{f_1}....p_k^{f_k}$ where the p_i are distinct primes. (We allow e_i or f_i to be zero)

$$Q(\xi_m) = Q(\xi_{p_1^{e_1}})Q(\xi_{p_2^{e_2}})...Q(\xi_{p_k^{e_k}})$$

and

$$Q(\xi_n) = Q(\xi_{p_1^{f_1}})Q(\xi_{p_2^{f_2}})...Q(\xi_{p_k^{f_k}})$$

Thus

$$Q(\xi_m, \xi_n) = Q(\xi_{p_1^{e_1}})........Q(\xi_{p_2^{e_k}})Q(\xi_{p_1^{f_1}})...Q(\xi_{p_k^{f_k}})$$
$$= Q(\xi_{p_1^{e_1}})Q(\xi_{p_1^{f_1}})...Q(\xi_{p_k^{e_k}})Q(\xi_{p_k^{f_k}})$$
$$= Q(\xi_{p_1^{\max(e_1,f_1)}})........Q(\xi_{p_1^{\max(e_k,f_k)}})$$
$$= Q(\xi_{p_1^{\max(e_1,f_1)}........p_1^{\max(e_k,f_k)}})$$
$$= Q(\xi_{[m,n]});$$

An entirely similar computation shows that
$$Q(\xi_m) \cap Q(\xi_n) = Q(\xi_{(m,n)})$$

Mutual information measures the information transferred when x_i is sent and y_i is received, and is defined as

$$I(x_i, y_i) = \log_2 \frac{P(x_i/y_i)}{P(x_i)} \ bits \qquad (1)$$

In a noise-free channel, **each** y_i is uniquely connected to the corresponding x_i, and so they constitute an input –output pair (x_i, y_i) for which

$$P(x_i/y_j) = 1 \ and \ I(x_i, y_j) = \log_2 \frac{1}{P(x_i)} \quad bits;$$

that is, the transferred information is equal to the self-information that corresponds to the input x_i In a very noisy channel, the output y_i and input x_i would be completely uncorrelated, and so $P(x_i/y_j) = P(x_i)$ and also $I(x_i, y_j) = 0$; that is, there is no transference of information. In general, a given channel will operate between these two extremes. The mutual information is defined between the input and the output of a given channel. An average of the calculation of the mutual information for all input-output pairs of a given channel is the average mutual information:

$$I(X,Y) = \sum_{i,j} P(x_i, y_j)I(x_i, y_j) = \sum_{i,j} P(x_i, y_j)\log_2\left[\frac{P(x_i/y_j)}{P(x_i)}\right]$$

bits per symbol . This calculation is done over the input and output alphabets. The average mutual information. The following expressions are useful for modifying the mutual information expression:

$$P(x_i, y_j) = P(x_i/y_j)P(y_j) = P(y_j/x_i)P(x_i)$$

$$P(y_j) = \sum_i P(y_j/x_i)P(x_i)$$

$$P(x_i) = \sum_i P(x_i/y_j)P(y_j)$$

Then

$$I(X,Y) = \sum_{i,j} P(x_i, y_j)$$

$$= \sum_{i,j} P(x_i, y_j)\log_2\left[\frac{1}{P(x_i)}\right]$$

$$-\sum_{i,j} P(x_i, y_j)\log_2\left[\frac{1}{P(x_i/y_j)}\right]$$

$$\sum_{i,j} P(x_i, y_j)\log_2\left[\frac{1}{P(x_i)}\right]$$

$$= \sum_i \left[P(x_i/y_j)P(y_j)\right]\log_2\frac{1}{P(x_i)}$$

$$\sum_i P(x_i)\log_2\frac{1}{P(x_i)} = H(X)$$

$$I(X,Y) = H(X) - H(X/Y)$$

Where $H(X/Y) = \sum_{i,j} P(x_i, y_j)\log_2\frac{1}{P(x_i/y_j)}$

is usually called the equivocation. In a sense, the equivocation can be seen as the information lost in the noisy channel, and is a function of the backward conditional probability. The observation of an output symbol y_j provides $H(X) - H(X/Y)$ bits of information. This difference is the mutual information of the channel. *Mutual Information: Properties* Since

$$P(x_i/y_j)P(y_j) = P(y_j/x_i)P(x_i)$$

The mutual information fits the condition
$$I(X,Y) = I(Y,X)$$

And by interchanging input and output it is also true that

$$I(X,Y) = H(Y) - H(Y/X)$$

Where

$$H(Y) = \sum_j P(y_j) \log_2 \frac{1}{P(y_j)}$$

This last entropy is usually called the noise entropy. Thus, the information transferred through the channel is the difference between the output entropy and the noise entropy. Alternatively, it can be said that the channel mutual information is the difference between the number of bits needed for determining a given input symbol before knowing the corresponding output symbol, and the number of bits needed for determining a given input symbol after knowing the corresponding output symbol

$$I(X,Y) = H(X) - H(X/Y)$$

As the channel mutual information expression is a difference between two quantities, it seems that this parameter can adopt negative values. However, and is spite of the fact that for some $y_j, H(X/y_j)$ can be larger than $H(X)$, this is not possible for the average value calculated over all the outputs:

$$\sum_{i,j} P(x_i, y_j) \log_2 \frac{P(x_i/y_j)}{P(x_i)} = \sum_{i,j} P(x_i, y_j) \log_2 \frac{P(x_i, y_j)}{P(x_i)P(y_j)}$$

Then

$$-I(X,Y) = \sum_{i,j} P(x_i, y_j) \frac{P(x_i)P(y_j)}{P(x_i, y_j)} \le 0$$

Because this expression is of the form

$$\sum_{i=1}^{M} P_i \log_2 (\frac{Q_i}{P_i}) \le 0$$

The above expression can be applied due to the factor $P(x_i)P(y_j)$, which is the product of two probabilities, so that it behaves as the quantity Q_i, which in this expression is a dummy variable that fits the condition $\sum_i Q_i \le 1$. It can be concluded that the average mutual information is a non-negative number. It can also be equal to zero, when the input and the output are independent of each other. A related entropy called the joint entropy is defined as

$$H(X,Y) = \sum_{i,j} P(x_i, y_j) \log_2 \frac{1}{P(x_i, y_j)}$$

$$= \sum_{i,j} P(x_i, y_j) \log_2 \frac{P(x_i)P(y_j)}{P(x_i, y_j)}$$

$$+ \sum_{i,j} P(x_i, y_j) \log_2 \frac{1}{P(x_i)P(y_j)}$$

Theorem 1.5: Entropies of the binary erasure channel (BEC) The BEC is defined with an alphabet of two inputs and three outputs, with symbol probabilities.

$$P(x_1) = \alpha \quad and \quad P(x_2) = 1 - \alpha, \quad and \quad transition$$

probabilities

$$P(y_3/x_2) = 1 - p \quad and \quad P(y_2/x_1) = 0,$$

$$and \quad P(y_3/x_1) = 0$$

$$and \quad P(y_1/x_2) = p$$

$$and \quad P(y_3/x_2) = 1 - p$$

Lemma 1.7. Given an arbitrary restricted time-discrete, amplitude-continuous channel whose restrictions are determined by sets F_n and whose density functions exhibit no dependence on the state s, let n be a fixed positive integer, and $p(x)$ an arbitrary probability density function on Euclidean n-space. $p(y|x)$ for the density $p_n(y_1,...,y_n | x_1,...x_n)$ and F for F_n. For any real number a, let

$$A = \left\{ (x,y) : \log \frac{p(y|x)}{p(y)} > a \right\} \quad (1)$$

Then for each positive integer u, there is a code (u, n, λ) such that

$$\lambda \le ue^{-a} + P\{(X,Y) \notin A\} + P\{X \notin F\} \quad (2)$$

Where

$$P\{(X,Y) \in A\} = \int_A ... \int p(x,y) dx dy, \qquad p(x,y) = p(x)p(y|x)$$

and

$$P\{X \in F\} = \int_F ... \int p(x) dx$$

Proof: A sequence $x^{(1)} \in F$ such that

$$P\{Y \in A_{x^1} | X = x^{(1)}\} \ge 1 - \varepsilon$$

$$where \quad A_x = \{ y : (x,y) \varepsilon A \};$$

Choose the decoding set B_1 to be $A_{x^{(1)}}$. Having chosen $x^{(1)},........,x^{(k-1)}$ and $B_1,...,B_{k-1}$, select $x^k \in F$ such that

$$P\left\{ Y \in A_{x^{(k)}} - \bigcup_{i=1}^{k-1} B_i | X = x^{(k)} \right\} \ge 1 - \varepsilon;$$

Set $B_k = A_{x^{(k)}} - \bigcup_{i=1}^{k-1} B_i$, If the process does not terminate in a finite number of steps, then the sequences $x^{(i)}$ and decoding sets $B_i, i = 1, 2, ..., u,$ form the desired code. Thus assume that the process

terminates after t steps. (Conceivably $t = 0$). We will show $t \geq u$ by showing that $\varepsilon \leq te^{-a} + P\{(X,Y) \notin A\} + P\{X \notin F\}$. We proceed as follows.
Let

$$B = \bigcup_{j=1}^{t} B_j. \quad (If \ t = 0, \ take \ B = \phi). \ Then$$

$$P\{(X,Y) \in A\} = \int_{(x,y) \in A} p(x,y)dx\,dy$$

$$= \int_x p(x) \int_{y \in A_x} p(y\,|\,x)dy\,dx$$

$$= \int_x p(x) \int_{y \in B \cap A_x} p(y\,|\,x)dy\,dx + \int_x p(x)$$

E. Algorithms

Ideals. Let A be a ring. Recall that an *ideal a* in A is a subset such that a is subgroup of A regarded as a group under addition;

$$a \in a, r \in A \Rightarrow ra \in A$$

The ideal generated by a subset S of A is the intersection of all ideals A containing a ----- it is easy to verify that this is in fact an ideal, and that it consist of all finite sums of the form $\sum r_i s_i$ with $r_i \in A, s_i \in S$. When $S = \{s_1, \ldots, s_m\}$, we shall write $(s_1, \ldots, s_m)$ for the ideal it generates.

Let a and b be ideals in A. The set $\{a + b \,|\, a \in a, b \in b\}$ is an ideal, denoted by $a + b$. The ideal generated by $\{ab \,|\, a \in a, b \in b\}$ is denoted by ab . Note that $ab \subset a \cap b$. Clearly ab consists of all finite sums $\sum a_i b_i$ with $a_i \in a$ and $b_i \in b$, and if $a = (a_1, \ldots, a_m)$ and $b = (b_1, \ldots, b_n)$, then $ab = (a_1 b_1, \ldots, a_i b_j, \ldots, a_m b_n)$ .Let a be an ideal of A. The set of cosets of a in A forms a ring A/a , and $a \mapsto a + a$ is a homomorphism $\phi: A \mapsto A/a$. The map $b \mapsto \phi^{-1}(b)$ is a one to one correspondence between the ideals of A/a and the ideals of A containing a An ideal p if *prime* if $p \neq A$ and $ab \in p \Rightarrow a \in p$ or $b \in p$. Thus p is prime if and only if A/p is nonzero and has the property that $ab = 0, \quad b \neq 0 \Rightarrow a = 0,$ i.e., A/p is an integral domain. An ideal m is *maximal* if $m \neq| A$ and there does not exist an ideal n contained strictly between m and A . Thus m is maximal if and only if A/m has no proper nonzero

ideals, and so is a field. Note that m maximal $\Rightarrow$ m prime. The ideals of $A \times B$ are all of the form $a \times b$, with a and b ideals in A and B . To see this, note that if c is an ideal in $A \times B$ and $(a,b) \in c$, then $(a,0) = (a,b)(1,0) \in c$ and $(0,b) = (a,b)(0,1) \in c$. This shows that $c = a \times b$ with
$$a = \{a \,|\, (a,b) \in c \ some \ b \in b\}$$
and
$$b = \{b \,|\, (a,b) \in c \ some \ a \in a\}$$

Let A be a ring. An A-algebra is a ring B together with a homomorphism $i_B : A \to B$. A *homomorphism of A*-algebra $B \to C$ is a homomorphism of rings $\varphi: B \to C$ such that $\varphi(i_B(a)) = i_C(a)$ for all $a \in A$. An A-algebra B is said to be *finitely generated* (or of *finite-type* over A) if there exist elements $x_1, \ldots, x_n \in B$ such that every element of B can be expressed as a polynomial in the x_i with coefficients in $i(A)$, i.e., such that the homomorphism $A[X_1, \ldots, X_n] \to B$ sending X_i to x_i is surjective. A ring homomorphism $A \to B$ is *finite,* and B is finitely generated as an A-module. Let k be a field, and let A be a k-algebra. If $1 \neq 0$ in A , then the map $k \to A$ is injective, we can identify k with its image, i.e., we can regard k as a subring of A . If 1=0 in a ring R, the R is the zero ring, i.e., $R = \{0\}$. **Polynomial rings.** Let k be a field. A *monomial* in $X_1, \ldots, X_n$ is an expression of the form $X_1^{a_1} \ldots X_n^{a_n},$ $a_j \in N$. The *total degree* of the monomial is $\sum a_i$. We sometimes abbreviate it by $X^\alpha, \alpha = (a_1, \ldots, a_n) \in \square^n$ The elements of the polynomial ring $k[X_1, \ldots, X_n]$ are finite sums $\sum c_{a_1 \ldots a_n} X_1^{a_1} \ldots X_n^{a_n},$ $c_{a_1 \ldots a_n} \in k,$ $a_j \in \square$ With the obvious notions of equality, addition and multiplication. Thus the monomials from basis for $k[X_1, \ldots, X_n]$ as a k-vector space. The ring $k[X_1, \ldots, X_n]$ is an integral domain, and the only units in it are the nonzero constant polynomials. A polynomial $f(X_1, \ldots, X_n)$ is *irreducible* if it is nonconstant and has only the obvious factorizations, i.e., $f = gh \Rightarrow g$ or h is constant. **Division in**

$k[X]$. The division algorithm allows us to divide a nonzero polynomial into another: let f and g be polynomials in $k[X]$ with $g \neq 0$; then there exist unique polynomials $q, r \in k[X]$ such that $f = qg + r$ with either $r = 0$ or $\deg r < \deg g$. Moreover, there is an algorithm for deciding whether $f \in (g)$, namely, find r and check whether it is zero. Moreover, the Euclidean algorithm allows to pass from finite set of generators for an ideal in $k[X]$ to a single generator by successively replacing each pair of generators with their greatest common divisor.

(Pure) lexicographic ordering (lex). Here monomials are ordered by lexicographic(dictionary) order. More precisely, let $\alpha = (a_1, \dots a_n)$ and $\beta = (b_1, \dots b_n)$ be two elements of $\square^n$; then $\alpha > \beta$ and $X^\alpha > X^\beta$ (lexicographic ordering) if, in the vector difference $\alpha - \beta \in \square$, the left most nonzero entry is positive. For example,

$XY^2 > Y^3Z^4$; $X^3Y^2Z^4 > X^3Y^2Z$. Note that this isn't quite how the dictionary would order them: it would put $XXXYYZZZZ$ after $XXXYYZ$. *Graded reverse lexicographic order (grevlex).* Here monomials are ordered by total degree, with ties broken by reverse lexicographic ordering. Thus, $\alpha > \beta$ if $\sum a_i > \sum b_i$, or $\sum a_i = \sum b_i$ and in $\alpha - \beta$ the right most nonzero entry is negative. For example:

$X^4Y^4Z^7 > X^5Y^5Z^4$ *(total degree greater)*
$XY^5Z^2 > X^4YZ^3$, $X^5YZ > X^4YZ^2$.

Orderings on $k[X_1, \dots X_n]$. Fix an ordering on the monomials in $k[X_1, \dots X_n]$. Then we can write an element f of $k[X_1, \dots X_n]$ in a canonical fashion, by re-ordering its elements in decreasing order. For example, we would write

$f = 4XY^2Z + 4Z^2 - 5X^3 + 7X^2Z^2$

as

$f = -5X^3 + 7X^2Z^2 + 4XY^2Z + 4Z^2$ *(lex)*

or

$f = 4XY^2Z + 7X^2Z^2 - 5X^3 + 4Z^2$ *(grevlex)*

Let $\sum a_\alpha X^\alpha \in k[X_1, \dots, X_n]$, in decreasing order:

$f = a_{\alpha_0} X^{\alpha_0} +_{\alpha_1} X^{\alpha_1} + \dots, \qquad \alpha_0 > \alpha_1 > \dots, \quad \alpha_0 \neq 0$

Then we define.

- The *multidegree* of f to be multdeg$(f) = \alpha_0$;
- The *leading coefficient of* f to be LC$(f) = a_{\alpha_0}$;
- The *leading monomial of* f to be LM$(f) = X^{\alpha_0}$;
- The *leading term of* f to be LT$(f) = a_{\alpha_0} X^{\alpha_0}$

For the polynomial $f = 4XY^2Z + \dots$, the multidegree is (1,2,1), the leading coefficient is 4, the leading monomial is XY^2Z, and the leading term is $4XY^2Z$. **The division algorithm in** $k[X_1, \dots X_n]$. Fix a monomial ordering in $\square^2$. Suppose given a polynomial f and an ordered set $(g_1, \dots g_s)$ of polynomials; the division algorithm then constructs polynomials $a_1, \dots a_s$ and r such that $f = a_1 g_1 + \dots + a_s g_s + r$ Where either $r = 0$ or no monomial in r is divisible by any of $LT(g_1), \dots, LT(g_s)$ **Step 1:** If $LT(g_1) | LT(f)$, divide g_1 into f to get

$$f = a_1 g_1 + h, \qquad a_1 = \frac{LT(f)}{LT(g_1)} \in k[X_1, \dots, X_n]$$

If $LT(g_1) | LT(h)$, repeat the process until $f = a_1 g_1 + f_1$ (different a_1) with $LT(f_1)$ not divisible by $LT(g_1)$. Now divide g_2 into f_1, and so on, until $f = a_1 g_1 + \dots + a_s g_s + r_1$ With $LT(r_1)$ not divisible by any $LT(g_1), \dots LT(g_s)$

Step 2: Rewrite $r_1 = LT(r_1) + r_2$, and repeat Step 1 with r_2 for f:

$f = a_1 g_1 + \dots + a_s g_s + LT(r_1) + r_3$ (different $a_i's$) **Monomial ideals.** In general, an ideal a will contain a polynomial without containing the individual terms of the polynomial; for example, the ideal $a = (Y^2 - X^3)$ contains $Y^2 - X^3$ but not Y^2 or X^3.

DEFINITION 1.5. An ideal a is *monomial* if
$$\sum c_\alpha X^\alpha \in a \Rightarrow X^\alpha \in a$$
all α with $c_\alpha \neq 0$.

PROPOSITION 1.3. Let a be a *monomial ideal,* and let $A = \{\alpha \mid X^\alpha \in a\}$. Then A satisfies the

condition $\alpha \in A$, $\beta \in \square^n \Rightarrow \alpha + \beta \in$ $(*)$

And a is the k -subspace of $k[X_1,...,X_n]$ generated by the $X^\alpha, \alpha \in A$. Conversely, of A is a subset of $\square^n$ satisfying $(*)$, then the k-subspace a of $k[X_1,...,X_n]$ generated by $\{X^\alpha \mid \alpha \in A\}$ is a monomial ideal.

PROOF. It is clear from its definition that a monomial ideal a is the k -subspace of $k[X_1,...,X_n]$ generated by the set of monomials it contains. If $X^\alpha \in a$ and $X^\beta \in k[X_1,...,X_n]$.

If a permutation is chosen uniformly and at random from the $n!$ possible permutations in S_n, then the counts $C_j^{(n)}$ of cycles of length j are dependent random variables. The joint distribution of $C^{(n)} = (C_1^{(n)},...,C_n^{(n)})$ follows from Cauchy's formula, and is given by

$$P[C^{(n)} = c] = \frac{1}{n!} N(n,c) = 1\left\{\sum_{j=1}^n jc_j = n\right\} \prod_{j=1}^n (\frac{1}{j})^{c_j} \frac{1}{c_j!}, \quad (1.1)$$

for $c \in \square_+^n$.

Lemma1.7 For nonnegative integers $m_1,...,m_n$,

$$E\left(\prod_{j=1}^n (C_j^{(n)})^{[m_j]}\right) = \left(\prod_{j=1}^n \left(\frac{1}{j}\right)^{m_j}\right) 1\left\{\sum_{j=1}^n jm_j \le n\right\} \quad (1.4)$$

Proof. This can be established directly by exploiting cancellation of the form $c_j^{[m_j]} / c_j! = 1/(c_j - m_j)!$ when $c_j \ge m_j$, which occurs between the ingredients in Cauchy's formula and the falling factorials in the moments. Write $m = \sum jm_j$. Then, with the first sum indexed by $c = (c_1,...c_n) \in \square_+^n$ and the last sum indexed by $d = (d_1,...,d_n) \in \square_+^n$ via the correspondence $d_j = c_j - m_j$, we have

$$E\left(\prod_{j=1}^n (C_j^{(n)})^{[m_j]}\right) = \sum_c P[C^{(n)} = c] \prod_{j=1}^n (c_j)^{[m_j]}$$

$$= \sum_{c:c_j \ge m_j \text{ for all } j} 1\left\{\sum_{j=1}^n jc_j = n\right\} \prod_{j=1}^n \frac{(c_j)^{[m_j]}}{j^{c_j} c_j!}$$

$$= \prod_{j=1}^n \frac{1}{j^{m_j}} \sum_d 1\left\{\sum_{j=1}^n jd_j = n-m\right\} \prod_{j=1}^n \frac{1}{j^{d_j} (d_j)!}$$

This last sum simplifies to the indicator $1(m \le n)$, corresponding to the fact that if $n-m \ge 0$, then $d_j = 0$ for $j > n-m$, and a random permutation in S_{n-m} must have some cycle structure $(d_1,...,d_{n-m})$. The moments of $C_j^{(n)}$ follow immediately as

$$E(C_j^{(n)})^{[r]} = j^{-r} 1\{jr \le n\} \quad (1.2)$$

We note for future reference that (1.4) can also be written in the form

$$E\left(\prod_{j=1}^n (C_j^{(n)})^{[m_j]}\right) = E\left(\prod_{j=1}^n Z_j^{[m_j]}\right) 1\left\{\sum_{j=1}^n jm_j \le n\right\}, \quad (1.3)$$

Where the Z_j are independent Poisson-distribution random variables that satisfy $E(Z_j) = 1/j$

The marginal distribution of cycle counts provides a formula for the joint distribution of the cycle counts C_j^n, we find the distribution of C_j^n using a combinatorial approach combined with the inclusion-exclusion formula.

Lemma 1.8. For $1 \le j \le n$,

$$P[C_j^{(n)} = k] = \frac{j^{-k}}{k!} \sum_{l=0}^{[n/j]-k} (-1)^l \frac{j^{-l}}{l!} \quad (1.1)$$

Proof. Consider the set I of all possible cycles of length j, formed with elements chosen from $\{1,2,...n\}$, so that $|I| = n^{[j]/j}$. For each $\alpha \in I$, consider the "property" G_α of having α; that is, G_α is the set of permutations $\pi \in S_n$ such that α is one of the cycles of π. We then have $|G_\alpha| = (n-j)!$, since the elements of $\{1,2,...,n\}$ not in α must be permuted among themselves. To use the inclusion-exclusion formula we need to calculate the term S_r, which is the sum of the probabilities of the r -fold intersection of properties, summing over all sets of r distinct properties. There are two cases to consider. If the r properties are indexed by r cycles having no elements in common, then the intersection specifies how rj elements are

moved by the permutation, and there are $(n-rj)!1(rj \leq n)$ permutations in the intersection. There are $n^{[rj]}/(j^r r!)$ such intersections. For the other case, some two distinct properties name some element in common, so no permutation can have both these properties, and the r-fold intersection is empty. Thus

$$S_r = (n-rj)!1(rj \leq n)$$

$$\times \frac{n^{[rj]}}{j^r r!}\frac{1}{n!} = 1(rj \leq n)\frac{1}{j^r r!}$$

Finally, the inclusion-exclusion series for the number of permutations having exactly k properties is

$$\sum_{l \geq 0}(-1)^l \binom{k+l}{l}S_{k+l},$$

Which simplifies to (1.1) Returning to the original hat-check problem, we substitute j=1 in (1.1) to obtain the distribution of the number of fixed points of a random permutation. For $k = 0,1,...,n$,

$$P[C_1^{(n)} = k] = \frac{1}{k!}\sum_{l=0}^{n-k}(-1)^l \frac{1}{l!}, \qquad (1.2)$$

and the moments of $C_1^{(n)}$ follow from (1.2) with $j = 1$. In particular, for $n \geq 2$, the mean and variance of $C_1^{(n)}$ are both equal to 1. The joint distribution of $(C_1^{(n)},...,C_b^{(n)})$ for any $1 \leq b \leq n$ has an expression similar to (1.7); this too can be derived by inclusion-exclusion. For any $c = (c_1,...,c_b) \in \Box_+^b$ with $m = \sum ic_i$,

$$P[(C_1^{(n)},...,C_b^{(n)}) = c]$$

$$= \left\{\prod_{i=1}^b \left(\frac{1}{i}\right)^{c_i}\frac{1}{c_i!}\right\}\sum_{\substack{l \geq 0 \text{ with} \\ \sum il_i \leq n-m}}(-1)^{l_1+...+l_b}\prod_{i=1}^b\left(\frac{1}{i}\right)^{l_i}\frac{1}{l_i!} \qquad (1.3)$$

The joint moments of the first b counts $C_1^{(n)},...,C_b^{(n)}$ can be obtained directly from (1.2) and (1.3) by setting $m_{b+1} = ... = m_n = 0$

The limit distribution of cycle counts
It follows immediately from Lemma 1.2 that for each fixed j, as $n \rightarrow \infty$,

$$P[C_j^{(n)} = k] \rightarrow \frac{j^{-k}}{k!}e^{-1/j}, \quad k = 0,1,2,...,$$

So that $C_j^{(n)}$ converges in distribution to a random variable Z_j having a Poisson distribution with mean $1/j$; we use the notation $C_j^{(n)} \rightarrow_d Z_j$

where $Z_j \Box P_o(1/j)$ to describe this. Infact, the limit random variables are independent.

Theorem 1.6 The process of cycle counts converges in distribution to a Poisson process of $\Box$ with intensity j^{-1}. That is, as $n \rightarrow \infty$,

$$(C_1^{(n)}, C_2^{(n)},...) \rightarrow_d (Z_1, Z_2,...) \qquad (1.1)$$

Where the $Z_j, j = 1,2,...$, are independent Poisson-distributed random variables with

$$E(Z_j) = \frac{1}{j}$$

Proof. To establish the converges in distribution one shows that for each fixed $b \geq 1$, as $n \rightarrow \infty$,

$$P[(C_1^{(n)},...,C_b^{(n)}) = c] \rightarrow P[(Z_1,...,Z_b) = c]$$

Error rates
The proof of Theorem says nothing about the rate of convergence. Elementary analysis can be used to estimate this rate when $b = 1$. Using properties of alternating series with decreasing terms, for $k = 0,1,...,n$,

$$\frac{1}{k!}\left(\frac{1}{(n-k+1)!} - \frac{1}{(n-k+2)!}\right) \leq \left|P[C_1^{(n)} = k] - P[Z_1 = k]\right|$$

$$\leq \frac{1}{k!(n-k+1)!}$$

It follows that

$$\frac{2^{n+1}}{(n+1)!}\frac{n}{n+2} \leq \sum_{k=0}^n \left|P[C_1^{(n)} = k] - P[Z_1 = k]\right| \leq \frac{2^{n+1}-1}{(n+1)!} \quad (1.11)$$

Since

$$P[Z_1 > n] = \frac{e^{-1}}{(n+1)!}\left(1 + \frac{1}{n+2} + \frac{1}{(n+2)(n+3)} + ...\right) < \frac{1}{(n+1)!},$$

We see from (1.11) that the total variation distance between the distribution $L(C_1^{(n)})$ of $C_1^{(n)}$ and the distribution $L(Z_1)$ of Z_1

Establish the asymptotics of $P\left[A_n(C^{(n)})\right]$ under conditions (A_0) and (B_{01}), where

$$A_n(C^{(n)}) = \bigcap_{1 \leq i \leq n}\bigcap_{r_i'+1 \leq j \leq r_i}\left\{C_{ij}^{(n)} = 0\right\},$$

and $\zeta_i = (r_i'/r_{id}) - 1 = O(i^{-g'})$ as $i \rightarrow \infty$, for some $g' > 0$. We start with the expression

$$P[A_n(C^{(n)})] = \frac{P[T_{0m}(Z^{'}) = n]}{P[T_{0m}(Z) = n]}$$

$$\prod_{\substack{1 \le i \le n \\ r_i^{'}+1 \le j \le r_i}} \left\{ 1 - \frac{\theta}{ir_i}(1 + E_{i0}) \right\} \qquad (1.1)$$

$$P[T_{0n}(Z^{'}) = n]$$

$$= \frac{\theta d}{n} \exp\left\{ \sum_{i \ge 1} [\log(1 + i^{-1}\theta d) - i^{-1}\theta d] \right\}$$

$$\left\{ 1 + O(n^{-1}\varphi_{\{1,2,7\}}^{'}(n)) \right\} \qquad (1.2)$$

and

$$P[T_{0n}(Z^{'}) = n]$$

$$= \frac{\theta d}{n} \exp\left\{ \sum_{i \ge 1} [\log(1 + i^{-1}\theta d) - i^{-1}\theta d] \right\}$$

$$\left\{ 1 + O(n^{-1}\varphi_{\{1,2,7\}}(n)) \right\} \qquad (1.3)$$

Where $\varphi_{\{1,2,7\}}^{'}(n)$ refers to the quantity derived from $Z^{'}$. It thus follows that $P[A_n(C^{(n)})] \square Kn^{-\theta(1-d)}$ for a constant K, depending on Z and the $r_i^{'}$ and computable explicitly from (1.1) – (1.3), if Conditions (A_0) and (B_{01}) are satisfied and if $\zeta_i^{*} = O(i^{-g^{'}})$ from some $g^{'} > 0$, since, under these circumstances, both $n^{-1}\varphi_{\{1,2,7\}}^{'}(n)$ and $n^{-1}\varphi_{\{1,2,7\}}(n)$ tend to zero as $n \to \infty$. In particular, for polynomials and square free polynomials, the relative error in this asymptotic approximation is of order n^{-1} if $g^{'} > 1$.

For $0 \le b \le n/8$ and $n \ge n_0$, with n_0

$$d_{TV}(L(C[1,b]), L(Z[1,b]))$$

$$\le d_{TV}(L(C[1,b]), L(Z[1,b]))$$

$$\le \varepsilon_{\{7,7\}}(n,b),$$

Where $\varepsilon_{\{7,7\}}(n,b) = O(b/n)$ under Conditions $(A_0), (D_1)$ and (B_{11}) Since, by the Conditioning Relation,

$$L(C[1,b] \mid T_{0b}(C) = l) = L(Z[1,b] \mid T_{0b}(Z) = l),$$

It follows by direct calculation that

$$d_{TV}(L(C[1,b]), L(Z[1,b]))$$

$$= d_{TV}(L(T_{0b}(C)), L(T_{0b}(Z)))$$

$$= \max_A \sum_{r \in A} P[T_{0b}(Z) = r]$$

$$\left\{ 1 - \frac{P[T_{bn}(Z) = n - r]}{P[T_{0n}(Z) = n]} \right\} \qquad (1.4)$$

Suppressing the argument Z from now on, we thus obtain

$$d_{TV}(L(C[1,b]), L(Z[1,b]))$$

$$= \sum_{r \ge 0} P[T_{0b} = r] \left\{ 1 - \frac{P[T_{bn} = n - r]}{P[T_{0n} = n]} \right\}_+$$

$$\le \sum_{r > n/2} P[T_{0b} = r] + \sum_{r=0}^{[n/2]} \frac{P[T_{0b} = r]}{P[T_{0b} = n]}$$

$$\times \left\{ \sum_{s=0}^{n} P[T_{0b} = s](P[T_{bn} = n - s] - P[T_{bn} = n - r] \right\}_+$$

$$\le \sum_{r > n/2} P[T_{0b} = r] + \sum_{r=0}^{[n/2]} P[T_{0b} = r]$$

$$\times \sum_{s=0}^{[n/2]} P[T_{0b} = s] \frac{\{P[T_{bn} = n - s] - P[T_{bn} = n - r]\}}{P[T_{0n} = n]}$$

$$+ \sum_{s=0}^{[n/2]} P[T_{0b} = r] \sum_{s=[n/2]+1}^{n} P[T = s] P[T_{bn} = n - s] / P[T_{0n} = n]$$

The first sum is at most $2n^{-1}ET_{0b}$; the third is bound by

$$(\max_{n/2 < s \le n} P[T_{0b} = s]) / P[T_{0n} = n]$$

$$\le \frac{2\varepsilon_{\{10.5(1)\}}(n/2, b)}{n} \frac{3n}{\theta P_\theta[0,1]},$$

$$\frac{3n}{\theta P_\theta[0,1]} 4n^{-2}\phi_{\{10.8\}}^{*}(n) \sum_{r=0}^{[n/2]} P[T_{0b} = r] \sum_{s=0}^{[n/2]} P[T_{0b} = s] \frac{1}{2}|r - s|$$

$$\le \frac{12\phi_{\{10.8\}}^{*}(n)}{\theta P_\theta[0,1]} \frac{ET_{0b}}{n}$$

Hence we may take

$$\varepsilon_{\{7,7\}}(n,b) = 2n^{-1}ET_{0b}(Z) \left\{ 1 + \frac{6\phi_{\{10.8\}}^{*}(n)}{\theta P_\theta[0,1]} \right\} P$$

$$+ \frac{6}{\theta P_\theta[0,1]} \varepsilon_{\{10.5(1)\}}(n/2, b) \qquad (1.5)$$

Required order under Conditions $(A_0), (D_1)$ and (B_{11}), if $S(\infty) < \infty$. If not, $\phi_{\{10.8\}}^{*}(n)$ can be

replaced by $\phi^*_{\{10.11\}}(n)$ in the above, which has the required order, without the restriction on the r_i implied by $S(\infty) < \infty$. Examining the Conditions $(A_0), (D_1)$ and (B_{11}), it is perhaps surprising to find that (B_{11}) is required instead of just (B_{01}); that is, that we should need $\sum_{l \geq 2} l \varepsilon_{il} = O(i^{-a_1})$ to hold for some $a_1 > 1$. A first observation is that a similar problem arises with the rate of decay of ε_{i1} as well. For this reason, n_1 is replaced by $\overset{\square}{n_1}$. This makes it possible to replace condition (A_1) by the weaker pair of conditions (A_0) and (D_1) in the eventual assumptions needed for $\varepsilon_{\{7,7\}}(n,b)$ to be of order $O(b/n)$; the decay rate requirement of order $i^{-1-\gamma}$ is shifted from ε_{i1} itself to its first difference. This is needed to obtain the right approximation error for the random mappings example. However, since all the classical applications make far more stringent assumptions about the $\varepsilon_{i1}, l \geq 2$, than are made in (B_{11}). The critical point of the proof is seen where the initial estimate of the difference $P[T^{(m)}_{bn} = s] - P[T^{(m)}_{bn} = s+1]$. The factor $\varepsilon_{\{10.10\}}(n)$, which should be small, contains a far tail element from $\overset{\square}{n_1}$ of the form $\phi^\theta_1(n) + u^*_1(n)$, which is only small if $a_1 > 1$, being otherwise of order $O(n^{1-a_1+\delta})$ for any $\delta > 0$, since $a_2 > 1$ is in any case assumed. For $s \geq n/2$, this gives rise to a contribution of order $O(n^{-1-a_1+\delta})$ in the estimate of the difference $P[T_{bn} = s] - P[T_{bn} = s+1]$, which, in the remainder of the proof, is translated into a contribution of order $O(tn^{-1-a_1+\delta})$ for differences of the form $P[T_{bn} = s] - P[T_{bn} = s+1]$, finally leading to a contribution of order $bn^{-a_1+\delta}$ for any $\delta > 0$ in $\varepsilon_{\{7,7\}}(n,b)$. Some improvement would seem to be possible, defining the function g by $g(w) = 1_{\{w=s\}} - 1_{\{w=s+t\}}$, differences that are of the form $P[T_{bn} = s] - P[T_{bn} = s+t]$ can be directly estimated, at a cost of only a single contribution of the form $\phi^\theta_1(n) + u^*_1(n)$. Then, iterating the cycle, in which one estimate of a

difference in point probabilities is improved to an estimate of smaller order, a bound of the form

$$\left| P[T_{bn} = s] - P[T_{bn} = s+t] \right| = O(n^{-2}t + n^{-1-a_1+\delta})$$

for any $\delta > 0$ could perhaps be attained, leading to a final error estimate in order $O(bn^{-1} + n^{-a_1+\delta})$ for any $\delta > 0$, to replace $\varepsilon_{\{7,7\}}(n,b)$. This would be of the ideal order $O(b/n)$ for large enough b, but would still be coarser for small b.

With b and n as in the previous section, we wish to show that

$$\left| d_{TV}(L(C[1,b]), L(Z[1,b])) - \frac{1}{2}(n+1)^{-1}|1-\theta| E \left| T_{0b} - ET_{0b} \right| \right|$$
$$\leq \varepsilon_{\{7,8\}}(n,b),$$

Where $\varepsilon_{\{7,8\}}(n,b) = O(n^{-1}b[n^{-1}b + n^{-\beta_{12}+\delta}])$ for any $\delta > 0$ under Conditions $(A_0), (D_1)$ and (B_{12}), with β_{12}. The proof uses sharper estimates. As before, we begin with the formula

$$d_{TV}(L(\overset{\square}{C}[1,b]), L(\overset{\square}{Z}[1,b]))$$
$$= \sum_{r \geq 0} P[T_{0b} = r] \left\{ 1 - \frac{P[T_{bn} = n-r]}{P[T_{0n} = n]} \right\}_+$$

Now we observe that

$$\left| \sum_{r \geq 0} P[T_{0b} = r] \left\{ 1 - \frac{P[T_{bn} = n-r]}{P[T_{0n} = n]} \right\}_+ - \sum_{r=0}^{[n/2]} \frac{P[T_{0b} = r]}{P[T_{0n} = n]} \right|$$
$$\times \left| \sum_{s=[n/2]+1}^{n} P[T_{0b} = s](P[T_{bn} = n-s] - P[T_{bn} = n-r]) \right|$$
$$\leq 4n^{-2}ET_{0b}^2 + (\max_{n/2 < s \leq n} P[T_{0b} = s]) / P[T_{0n} = n]$$
$$+ P[T_{0b} > n/2]$$
$$\leq 8n^{-2}ET_{0b}^2 + \frac{3\varepsilon_{\{10.5(2)\}}(n/2,b)}{\theta P_\theta[0,1]}, \qquad (1.1)$$

We have

$$\left| \sum_{r=0}^{[n/2]} \frac{P[T_{0b}=r]}{P[T_{0n}=n]} \right.$$

$$\times \left(\left\{ \sum_{s=0}^{[n/2]} P[T_{0b}=s](P[T_{bn}=n-s]-P[T_{bn}=n-r] \right\}_+ \right.$$

$$\left. - \left\{ \sum_{s=0}^{[n/2]} P[T_{0b}=s] \frac{(s-r)(1-\theta)}{n+1} P[T_{0n}=n] \right\}_+ \right) \left. \right|$$

$$\leq \frac{1}{n^2 P[T_{0n}=n]} \sum_{r\geq 0} P[T_{0b}=r] \sum_{s\geq 0} P[T_{0b}=s]|s-r|$$

$$\times \left\{ \varepsilon_{\{10.14\}}(n,b) + 2(r\vee s)|1-\theta|n^{-1}\left\{ K_0\theta + 4\phi^*_{\{10.8\}}(n) \right\} \right\}$$

$$\leq \frac{6}{\theta n P_\theta[0,1]} ET_{0b}\varepsilon_{\{10.14\}}(n,b)$$

$$+4|1-\theta|n^{-2}ET_{0b}^2\left\{ K_0\theta + 4\phi^*_{\{10.8\}}(n) \right\}$$

$$(\frac{3}{\theta n P_\theta[0,1]}) \left. \right\}, \qquad (1.2)$$

The approximation in (1.2) is further simplified by noting that

$$\sum_{r=0}^{[n/2]} P[T_{0b}=r] \left| \left\{ \sum_{s=0}^{[n/2]} P[T_{0b}=s]\frac{(s-r)(1-\theta)}{n+1} \right\}_+ \right.$$

$$\left. - \left\{ \sum_{s\geq 0} P[T_{0b}=s]\frac{(s-r)(1-\theta)}{n+1} \right\}_+ \right|$$

$$\leq \sum_{r=0}^{[n/2]} P[T_{0b}=r] \sum_{s>[n/2]} P[T_{0b}=s]\frac{(s-r)|1-\theta|}{n+1}$$

$$\leq |1-\theta|n^{-1}E(T_{0b}1\{T_{0b}>n/2\}) \leq 2|1-\theta|n^{-2}ET_{0b}^2, \qquad (1.3)$$

and then by observing that

$$\sum_{r>[n/2]} P[T_{0b}=r] \left\{ \sum_{s\geq 0} P[T_{0b}=s]\frac{(s-r)(1-\theta)}{n+1} \right\}$$

$$\leq n^{-1}|1-\theta|(ET_{0b}P[T_{0b}>n/2]+E(T_{0b}1\{T_{0b}>n/2\}))$$

$$\leq 4|1-\theta|n^{-2}ET_{0b}^2 \qquad (1.4)$$

Combining the contributions of (1.2) –(1.3), we thus find   tha

$$\left| d_{TV}(L(\overset{\square}{C}[1,b]),L(\overset{\square}{Z}[1,b])) \right.$$

$$\left. -(n+1)^{-1}\sum_{r\geq 0} P[T_{0b}=r]\left\{ \sum_{s\geq 0} P[T_{0b}=s](s-r)(1-\theta) \right\}_+ \right|$$

$$\leq \varepsilon_{\{7.8\}}(n,b)$$

$$= \frac{3}{\theta P_\theta[0,1]}\left\{ \varepsilon_{\{10.5(2)\}}(n/2,b) + 2n^{-1}ET_{0b}\varepsilon_{\{10.14\}}(n,b) \right\}$$

$$+ 2n^{-2}ET_{0b}^2\left\{ 4+3|1-\theta| + \frac{24|1-\theta|\phi^*_{\{10.8\}}(n)}{\theta P_\theta[0,1]} \right\} \qquad (1.5)$$

The quantity $\varepsilon_{\{7.8\}}(n,b)$ is seen to be of the order claimed under Conditions $(A_0),(D_1)$ and (B_{12}) , provided that $S(\infty)<\infty$; this supplementary condition can be removed if $\phi^*_{\{10.8\}}(n)$ is replaced by $\phi^*_{\{10.11\}}(n)$ in the definition of $\varepsilon_{\{7.8\}}(n,b)$, has the required order without the restriction on the r_i implied by assuming that $S(\infty)<\infty$. Finally, a direct calculation now shows that

$$\sum_{r\geq 0} P[T_{0b}=r]\left\{ \sum_{s\geq 0} P[T_{0b}=s](s-r)(1-\theta) \right\}_+$$

$$= \frac{1}{2}|1-\theta|E|T_{0b}-ET_{0b}|$$

Example 1.0. Consider the point $O=(0,...,0)\in \square^n$. For an arbitrary vector r , the coordinates of the point $x=O+r$ are equal to the respective coordinates of the vector $r: x=(x^1,...x^n)$ and $r=(x^1,...,x^n)$. The vector r such as in the example is called the position vector or the radius vector of the point x . (Or, in greater detail: r is the radius-vector of x w.r.t an origin O). Points are frequently specified by their radius-vectors. This presupposes the choice of O as the "standard origin". Let us summarize. We have considered $\square^n$ and interpreted its elements in two ways: as points and as vectors. Hence we may say that we leading with the two copies of $\square^n$: $\square^n =$ {points}, $\square^n=$ {vectors}

Operations with vectors: multiplication by a number, addition. Operations with points and vectors: adding a vector to a point (giving a point), subtracting two points (giving a vector). $\square^n$ treated in this way is called an *n-dimensional affine space*. *(*An "abstract" affine space is a pair of sets , the set of points and the set of vectors so that the operations as above are defined axiomatically). Notice that vectors in an affine space are also known as "free

vectors". Intuitively, they are not fixed at points and "float freely" in space. From $\square^n$ considered as an affine space we can precede in two opposite directions: $\square^n$ as an Euclidean space $\Leftarrow$ $\square^n$ as an affine space $\Rightarrow$ $\square^n$ as a manifold. Going to the left means introducing some extra structure which will make the geometry richer. Going to the right means forgetting about part of the affine structure; going further in this direction will lead us to the so-called "smooth (or differentiable) manifolds". The theory of differential forms does not require any extra geometry. So our natural direction is to the right. The Euclidean structure, however, is useful for examples and applications. So let us say a few words about it:

Remark 1.0. *Euclidean geometry.* In $\square^n$ considered as an affine space we can already do a good deal of geometry. For example, we can consider lines and planes, and quadric surfaces like an ellipsoid. However, we cannot discuss such things as "lengths", "angles" or "areas" and "volumes". To be able to do so, we have to introduce some more definitions, making $\square^n$ a Euclidean space. Namely, we define the length of a vector $a = (a^1,...,a^n)$ to be

$$|a| := \sqrt{(a^1)^2 + ... + (a^n)^2} \qquad (1)$$

After that we can also define distances between points as follows:

$$d(A,B) := \left| \overrightarrow{AB} \right| \qquad (2)$$

One can check that the distance so defined possesses natural properties that we expect: is it always non-negative and equals zero only for coinciding points; the distance from A to B is the same as that from B to A (symmetry); also, for three points, A, B and C, we have $d(A,B) \le d(A,C) + d(C,B)$ (the "triangle inequality"). To define angles, we first introduce the scalar product of two vectors

$$(a,b) := a^1 b^1 + ... + a^n b^n \qquad (3)$$

Thus $|a| = \sqrt{(a,a)}$. The scalar product is also denote by dot: $a.b = (a,b)$, and hence is often referred to as the "dot product". Now, for nonzero vectors, we define the angle between them by the equality

$$\cos \alpha := \frac{(a,b)}{|a||b|} \qquad (4)$$

The angle itself is defined up to an integral multiple of 2π. For this definition to be consistent we have to ensure that the r.h.s. of (4) does not exceed 1 by the absolute value. This follows from the inequality

$$(a,b)^2 \le |a|^2 |b|^2 \qquad (5)$$

known as the Cauchy–Bunyakovsky–Schwarz inequality (various combinations of these three names are applied in different books). One of the ways of proving (5) is to consider the scalar square of the linear combination $a + tb$, where $t \in R$. As $(a + tb, a + tb) \ge 0$ is a quadratic polynomial in t which is never negative, its discriminant must be less or equal zero. Writing this explicitly yields (5). The triangle inequality for distances also follows from the inequality (5).

Example 1.1. Consider the function $f(x) = x^i$ (the i-th coordinate). The linear function dx^i (the differential of x^i) applied to an arbitrary vector h is simply h^i. From these examples follows that we can rewrite df as

$$df = \frac{\partial f}{\partial x^1} dx^1 + ... + \frac{\partial f}{\partial x^n} dx^n, \qquad (1)$$

which is the standard form. Once again: the partial derivatives in (1) are just the coefficients (depending on x); $dx^1, dx^2,...$ are linear functions giving on an arbitrary vector h its coordinates $h^1, h^2,...,$ respectively. Hence

$$df(x)(h) = \partial_{hf(x)} = \frac{\partial f}{\partial x^1} h^1 +$$
$$... + \frac{\partial f}{\partial x^n} h^n, \qquad (2)$$

Theorem 1.7. Suppose we have a parametrized curve $t \mapsto x(t)$ passing through $x_0 \in \square^n$ at $t = t_0$ and with the velocity vector $x(t_0) = \upsilon$ Then

$$\frac{df(x(t))}{dt}(t_0) = \partial_\upsilon f(x_0) = df(x_0)(\upsilon) \qquad (1)$$

Proof. Indeed, consider a small increment of the parameter $t : t_0 \mapsto t_0 + \Delta t$, Where $\Delta t \mapsto 0$. On the other hand, we have $f(x_0 + h) - f(x_0) = df(x_0)(h) + \beta(h)|h|$ for an arbitrary vector h, where $\beta(h) \to 0$ when $h \to 0$. Combining it together, for the increment of $f(x(t))$ we obtain

$$f(x(t_0 + \Delta t) - f(x_0)$$
$$= df(x_0)(\upsilon . \Delta t + \alpha(\Delta t)\Delta t)$$
$$+ \beta(\upsilon . \Delta t + \alpha(\Delta t)\Delta t).\left|\upsilon \Delta t + \alpha(\Delta t)\Delta t\right|$$
$$= df(x_0)(\upsilon). \Delta t + \gamma(\Delta t)\Delta t$$

For a certain $\gamma(\Delta t)$ such that $\gamma(\Delta t) \to 0$ when $\Delta t \to 0$ (we used the linearity of $df(x_0)$). By the definition, this means that the derivative of $f(x(t))$ at $t = t_0$ is exactly $df(x_0)(\upsilon)$. The statement of the theorem can be expressed by a simple formula:

$$\frac{df(x(t))}{dt} = \frac{\partial f}{\partial x^1} x^1 + ... + \frac{\partial f}{\partial x^n} x^n \qquad (2)$$

To calculate the value Of df at a point x_0 on a given vector υ one can take an arbitrary curve passing Through x_0 at t_0 with υ as the velocity vector at t_0 and calculate the usual derivative of $f(x(t))$ at $t = t_0$.

Theorem 1.8. For functions $f, g : U \to \square$, $U \subset \square^n$,

$$d(f + g) = df + dg \qquad (1)$$
$$d(fg) = df . g + f . dg \qquad (2)$$

Proof. Consider an arbitrary point x_0 and an arbitrary vector υ stretching from it. Let a curve $x(t)$ be such that $x(t_0) = x_0$ and $x(t_0) = \upsilon$. Hence

$$d(f + g)(x_0)(\upsilon) = \frac{d}{dt}(f(x(t)) + g(x(t)))$$

at $t = t_0$ and

$$d(fg)(x_0)(\upsilon) = \frac{d}{dt}(f(x(t))g(x(t)))$$

at $t = t_0$ Formulae (1) and (2) then immediately follow from the corresponding formulae for the usual derivative Now, almost without change the theory generalizes to functions taking values in $\square^m$ instead of $\square$. The only difference is that now the differential of a map $F : U \to \square^m$ at a point x will be a linear function taking vectors in $\square^n$ to vectors in $\square^m$ (instead of $\square$). For an arbitrary vector $h \in |\square^n$,

$$F(x + h) = F(x) + dF(x)(h)$$
$$+ \beta(h)|h| \qquad (3)$$

Where $\beta(h) \to 0$ when $h \to 0$. We have $dF = (dF^1, ..., dF^m)$ and

$$dF = \frac{\partial F}{\partial x^1} dx^1 + ... + \frac{\partial F}{\partial x^n} dx^n$$

$$= \begin{pmatrix} \dfrac{\partial F^1}{\partial x^1} .... \dfrac{\partial F^1}{\partial x^n} \\ ... \quad ... \quad ... \\ \dfrac{\partial F^m}{\partial x^1} .... \dfrac{\partial F^m}{\partial x^n} \end{pmatrix} \begin{pmatrix} dx^1 \\ ... \\ dx^n \end{pmatrix} \qquad (4)$$

In this matrix notation we have to write vectors as vector-columns.

Theorem 1.9. For an arbitrary parametrized curve $x(t)$ in $\square^n$, the differential of a map $F : U \to \square^m$ (where $U \subset \square^n$) maps the velocity vector $x(t)$ to the velocity vector of the curve $F(x(t))$ in $\square^m$:

$$\frac{dF(x(t))}{dt} = dF(x(t))(\dot{x}(t)) \qquad (1)$$

Proof. By the definition of the velocity vector,

$$x(t + \Delta t) = x(t) + \dot{x}(t).\Delta t + \alpha(\Delta t)\Delta t \qquad (2)$$

Where $\alpha(\Delta t) \to 0$ when $\Delta t \to 0$. By the definition of the differential,

$$F(x + h) = F(x) + dF(x)(h) + \beta(h)|h| \qquad (3)$$

Where $\beta(h) \to 0$ when $h \to 0$. we obtain

$$F(x(t + \Delta t)) = F(x + \underbrace{\dot{x}(t).\Delta t + \alpha(\Delta t)\Delta t}_{h})$$

$$= F(x) + dF(x)(\dot{x}(t)\Delta t + \alpha(\Delta t)\Delta t) +$$

$$\beta(\dot{x}(t)\Delta t + \alpha(\Delta t)\Delta t).\left|\dot{x}(t)\Delta t + \alpha(\Delta t)\Delta t\right|$$

$$= F(x) + dF(x)(\dot{x}(t)\Delta t + \gamma(\Delta t)\Delta t$$

For some $\gamma(\Delta t) \to 0$ when $\Delta t \to 0$. This precisely means that $dF(x)\dot{x}(t)$ is the velocity vector of $F(x)$. As every vector attached to a point can be viewed as the velocity vector of some curve

passing through this point, this theorem gives a clear geometric picture of dF as a linear map on vectors.

Theorem 1.10 Suppose we have two maps $F : U \to V$ and $G : V \to W$, where $U \subset \square^n, V \subset \square^m, W \subset \square^p$ (open domains). Let $F : x \mapsto y = F(x)$. Then the differential of the composite map $GoF : U \to W$ is the composition of the differentials of F and G:

$$d(GoF)(x) = dG(y)odF(x) \qquad (4)$$

Proof. We can use the description of the differential .Consider a curve $x(t)$ in $\square^n$ with the velocity vector $\dot{x}$. Basically, we need to know to which vector in $\square^p$ it is taken by $d(GoF)$. the curve $(GoF)(x(t) = G(F(x(t))$. By the same theorem, it equals the image under dG of the Anycast Flow vector to the curve $F(x(t))$ in $\square^m$. Applying the theorem once again, we see that the velocity vector to the curve $F(x(t))$ is the image under dF of the vector $\dot{x}(t)$. Hence

$$d(GoF)(\dot{x}) = dG(dF(\dot{x})) \qquad \text{for an arbitrary}$$

vector $\dot{x}$.

Corollary 1.0. If we denote coordinates in $\square^n$ by $(x^1, ..., x^n)$ and in $\square^m$ by $(y^1, ..., y^m)$, and write

$$dF = \frac{\partial F}{\partial x^1} dx^1 + ... + \frac{\partial F}{\partial x^n} dx^n \qquad (1)$$

$$dG = \frac{\partial G}{\partial y^1} dy^1 + ... + \frac{\partial G}{\partial y^n} dy^n, \qquad (2)$$

Then the chain rule can be expressed as follows:

$$d(GoF) = \frac{\partial G}{\partial y^1} dF^1 + ... + \frac{\partial G}{\partial y^m} dF^m, \qquad (3)$$

Where dF^i are taken from (1). In other words, to get $d(GoF)$ we have to substitute into (2) the expression for $dy^i = dF^i$ from (3). This can also be expressed by the following matrix formula:

$$d(GoF) = \begin{pmatrix} \frac{\partial G^1}{\partial y^1} & .... & \frac{\partial G^1}{\partial y^m} \\ ... & ... & ... \\ \frac{\partial G^p}{\partial y^1} & ... & \frac{\partial G^p}{\partial y^m} \end{pmatrix} \begin{pmatrix} \frac{\partial F^1}{\partial x^1} & .... & \frac{\partial F^1}{\partial x^n} \\ ... & ... & ... \\ \frac{\partial F^m}{\partial x^1} & ... & \frac{\partial F^m}{\partial x^n} \end{pmatrix} \begin{pmatrix} dx^1 \\ ... \\ dx^n \end{pmatrix} \qquad (4)$$

i.e., if dG and dF are expressed by matrices of partial derivatives, then $d(GoF)$ is expressed by the product of these matrices. This is often written as

$$\begin{pmatrix} \frac{\partial z^1}{\partial x^1} & .... & \frac{\partial z^1}{\partial x^n} \\ ... & ... & ... \\ \frac{\partial z^p}{\partial x^1} & ... & \frac{\partial z^p}{\partial x^n} \end{pmatrix} = \begin{pmatrix} \frac{\partial z^1}{\partial y^1} & .... & \frac{\partial z^1}{\partial y^m} \\ ... & ... & ... \\ \frac{\partial z^p}{\partial y^1} & ... & \frac{\partial z^p}{\partial y^m} \end{pmatrix}$$

$$\begin{pmatrix} \frac{\partial y^1}{\partial x^1} & .... & \frac{\partial y^1}{\partial x^n} \\ ... & ... & ... \\ \frac{\partial y^m}{\partial x^1} & ... & \frac{\partial y^m}{\partial x^n} \end{pmatrix}, \qquad (5)$$

Or

$$\frac{\partial z^\mu}{\partial x^a} = \sum_{i=1}^{m} \frac{\partial z^\mu}{\partial y^i} \frac{\partial y^i}{\partial x^a}, \qquad (6)$$

Where it is assumed that the dependence of $y \in \square^m$ on $x \in \square^n$ is given by the map F, the dependence of $z \in \square^p$ on $y \in \square^m$ is given by the map G, and the dependence of $z \in \square^p$ on $x \in \square^n$ is given by the composition GoF.

Definition 1.6. Consider an open domain $U \subset \square^n$. Consider also another copy of $\square^n$, denoted for distinction $\square^n_y$, with the standard coordinates $(y^1 ... y^n)$. A system of coordinates in the open domain U is given by a map $F : V \to U$, where $V \subset \square^n_y$ is an open domain of $\square^n_y$, such that the following three conditions are satisfied :
(1) F is smooth;
(2) F is invertible;
(3) $F^{-1} : U \to V$ is also smooth

The coordinates of a point $x \in U$ in this system are the standard coordinates of $F^{-1}(x) \in \square^n_y$
In other words,

$$F : (y^1 ..., y^n) \mapsto x = x(y^1 ..., y^n) \qquad (1)$$

Here the variables $(y^1 ..., y^n)$ are the "new" coordinates of the point x

Example 1.2. Consider a curve in $\square^2$ specified in polar coordinates as

$$x(t): r = r(t), \varphi = \varphi(t) \qquad (1)$$

We can simply use the chain rule. The map $t \mapsto x(t)$ can be considered as the composition of the maps $t \mapsto (r(t), \varphi(t)), (r, \varphi) \mapsto x(r, \varphi)$. Then, by the chain rule, we have

$$\dot{x} = \frac{dx}{dt} = \frac{\partial x}{\partial r}\frac{dr}{dt} + \frac{\partial x}{\partial \varphi}\frac{d\varphi}{dt} = \frac{\partial x}{\partial r}\dot{r} + \frac{\partial x}{\partial \varphi}\dot{\varphi} \qquad (2)$$

Here $\dot{r}$ and $\dot{\varphi}$ are scalar coefficients depending on t, whence the partial derivatives $\partial x / \partial r, \partial x / \partial \varphi$ are vectors depending on point in $\Box^2$. We can compare this with the formula in the "standard" coordinates:

$\dot{x} = e_1 \dot{x} + e_2 \dot{y}$. Consider the vectors $\partial x / \partial r, \partial x / \partial \varphi$. Explicitly we have

$$\frac{\partial x}{\partial r} = (\cos \varphi, \sin \varphi) \qquad (3)$$

$$\frac{\partial x}{\partial \varphi} = (-r \sin \varphi, r \cos \varphi) \qquad (4)$$

From where it follows that these vectors make a basis at all points except for the origin (where $r = 0$). It is instructive to sketch a picture, drawing vectors corresponding to a point as starting from that point. Notice that $\partial x / \partial r, \partial x / \partial \varphi$ are, respectively, the velocity vectors for the curves $r \mapsto x(r, \varphi)$ ($\varphi = \varphi_0 \ fixed$) and $\varphi \mapsto x(r, \varphi)$ ($r = r_0 \ fixed$). We can conclude that for an arbitrary curve given in polar coordinates the velocity vector will have components $(\dot{r}, \dot{\varphi})$ if as a basis we take $e_r := \partial x / \partial r, e_\varphi := \partial x / \partial \varphi$:

$$\dot{x} = e_r \dot{r} + e_\varphi \dot{\varphi} \qquad (5)$$

A characteristic feature of the basis e_r, e_φ is that it is not "constant" but depends on point. Vectors "stuck to points" when we consider curvilinear coordinates.

Proposition 1.3. The velocity vector has the same appearance in all coordinate systems.
Proof. Follows directly from the chain rule and the transformation law for the basis e_i. In particular, the elements of the basis $e_i = \partial x / \partial x^i$ (originally, a formal notation) can be understood directly as the velocity vectors of the coordinate lines

$x^i \mapsto x(x^1, ..., x^n)$ (all coordinates but x^i are fixed). Since we now know how to handle velocities in arbitrary coordinates, the best way to treat the differential of a map $F: \Box^n \to \Box^m$ is by its action on the velocity vectors. By definition, we set

$$dF(x_0): \frac{dx(t)}{dt}(t_0) \mapsto \frac{dF(x(t))}{dt}(t_0) \qquad (1)$$

Now $dF(x_0)$ is a linear map that takes vectors attached to a point $x_0 \in \Box^n$ to vectors attached to the point $F(x) \in \Box^m$

$$dF = \frac{\partial F}{\partial x^1}dx^1 + ... + \frac{\partial F}{\partial x^n}dx^n$$

$$(e_1, ..., e_m)\begin{pmatrix} \frac{\partial F^1}{\partial x^1} \cdots \frac{\partial F^1}{\partial x^n} \\ ... \ ... \ ... \\ \frac{\partial F^m}{\partial x^1} \cdots \frac{\partial F^m}{\partial x^n} \end{pmatrix}\begin{pmatrix} dx^1 \\ ... \\ dx^n \end{pmatrix}, \qquad (2)$$

In particular, for the differential of a function we always have

$$df = \frac{\partial f}{\partial x^1}dx^1 + ... + \frac{\partial f}{\partial x^n}dx^n, \qquad (3)$$

Where x^i are arbitrary coordinates. The form of the differential does not change when we perform a change of coordinates.

Example 1.3 Consider a 1-form in $\Box^2$ given in the standard coordinates:

$A = -ydx + xdy$ In the polar coordinates we will have $x = r \cos \varphi, y = r \sin \varphi$, hence

$$dx = \cos \varphi dr - r \sin \varphi d\varphi$$

$$dy = \sin \varphi dr + r \cos \varphi d\varphi$$

Substituting into A, we get
$$A = -r \sin \varphi (\cos \varphi dr - r \sin \varphi d\varphi)$$

$$+r \cos \varphi (\sin \varphi dr + r \cos \varphi d\varphi)$$

$$= r^2 (\sin^2 \varphi + \cos^2 \varphi) d\varphi = r^2 d\varphi$$

Hence $A = r^2 d\varphi$ is the formula for A in the polar coordinates. In particular, we see that this is again a 1-form, a linear combination of the differentials of coordinates with functions as coefficients. Secondly, in a more conceptual way, we can define a 1-form in a domain U as a linear function on vectors at every point of U:

$$\omega(\upsilon) = \omega_1 \upsilon^1 + ... + \omega_n \upsilon^n, \qquad (1)$$

If $\upsilon = \sum e_i \upsilon^i$, where $e_i = \partial x \big/ \partial x^i$. Recall that the differentials of functions were defined as linear functions on vectors (at every point), and

$$dx^i(e_j) = dx^i\left(\frac{\partial x}{\partial x^j}\right) = \delta_j^i \qquad (2) \qquad \text{at}$$

every point x.

Theorem 1.9. For arbitrary 1-form ω and path γ, the integral $\int_\gamma \omega$ does not change if we change parametrization of γ provide the orientation remains the same.

Proof: Consider $\left\langle \omega(x(t)), \dfrac{dx}{dt^{'}} \right\rangle$ and

$\left\langle \omega(x(t(t^{'}))), \dfrac{dx}{dt^{'}} \right\rangle$ As

$\left\langle \omega(x(t(t^{'}))), \dfrac{dx}{dt^{'}} \right\rangle = \left\| \left\langle \omega(x(t(t^{'}))), \dfrac{dx}{dt^{'}} \right\rangle \cdot \dfrac{dt}{dt^{'}} \right.$,

Let p be a rational prime and let $K = \square(\zeta_p)$. We write ζ for ζ_p or this section. Recall that K has degree $\varphi(p) = p-1$ over $\square$. We wish to show that $O_K = \square[\zeta]$. Note that ζ is a root of $x^p - 1$, and thus is an algebraic integer; since O_K is a ring we have that $\square[\zeta] \subseteq O_K$. We give a proof without assuming unique factorization of ideals. We begin with some norm and trace computations. Let j be an integer. If j is not divisible by p, then ζ^j is a primitive p^{th} root of unity, and thus its conjugates are $\zeta, \zeta^2, ..., \zeta^{p-1}$. Therefore

$$Tr_{K/\square}(\zeta^j) = \zeta + \zeta^2 + ... + \zeta^{p-1} = \Phi_p(\zeta) - 1 = -1$$

If p does divide j, then $\zeta^j = 1$, so it has only the one conjugate 1, and $Tr_{K/\square}(\zeta^j) = p-1$ By linearity of the trace, we find that

$$Tr_{K/\square}(1-\zeta) = Tr_{K/\square}(1-\zeta^2) = ...$$
$$= Tr_{K/\square}(1-\zeta^{p-1}) = p$$

We also need to compute the norm of $1-\zeta$. For this, we use the factorization

$$x^{p-1} + x^{p-2} + ... + 1 = \Phi_p(x)$$
$$= (x-\zeta)(x-\zeta^2)...(x-\zeta^{p-1});$$

Plugging in $x = 1$ shows that

$$p = (1-\zeta)(1-\zeta^2)...(1-\zeta^{p-1})$$

Since the $(1-\zeta^j)$ are the conjugates of $(1-\zeta)$, this shows that $N_{K/\square}(1-\zeta) = p$ The key result for determining the ring of integers O_K is the following.

LEMMA 1.9

$$(1-\zeta)O_K \cap \square = p\square$$

Proof. We saw above that p is a multiple of $(1-\zeta)$ in O_K, so the inclusion $(1-\zeta)O_K \cap \square \supseteq p\square$ is immediate. Suppose now that the inclusion is strict. Since $(1-\zeta)O_K \cap \square$ is an ideal of $\square$ containing $p\square$ and $p\square$ is a maximal ideal of $\square$, we must have $(1-\zeta)O_K \cap \square = \square$ Thus we can write

$$1 = \alpha(1-\zeta)$$

For some $\alpha \in O_K$. That is, $1-\zeta$ is a unit in O_K.

COROLLARY 1.1 For any $\alpha \in O_K$, $Tr_{K/\square}((1-\zeta)\alpha) \in p.\square$

PROOF. We have

$$Tr_{K/\square}((1-\zeta)\alpha) = \sigma_1((1-\zeta)\alpha) + ... + \sigma_{p-1}((1-\zeta)\alpha)$$
$$= \sigma_1(1-\zeta)\sigma_1(\alpha) + ... + \sigma_{p-1}(1-\zeta)\sigma_{p-1}(\alpha)$$
$$= (1-\zeta)\sigma_1(\alpha) + ... + (1-\zeta^{p-1})\sigma_{p-1}(\alpha)$$

Where the σ_i are the complex embeddings of K (which we are really viewing as automorphisms of K) with the usual ordering. Furthermore, $1-\zeta^j$ is a multiple of $1-\zeta$ in O_K for every $j \neq 0$. Thus $Tr_{K/\square}(\alpha(1-\zeta)) \in (1-\zeta)O_K$ Since the trace is also a rational integer.

PROPOSITION 1.4 Let p be a prime number and let $K = |\square(\zeta_p)$ be the pth cyclotomic field. Then $O_K = \square[\zeta_p] \cong \square[x]/(\Phi_p(x));$ Thus $1, \zeta_p, ..., \zeta_p^{p-2}$ is an integral basis for O_K.

PROOF. Let $\alpha \in O_K$ and write

$$\alpha = a_0 + a_1\zeta + ... + a_{p-2}\zeta^{p-2} \qquad \text{With} \quad a_i \in \square.$$

Then

$$\alpha(1-\zeta) = a_0(1-\zeta) + a_1(\zeta - \zeta^2) + ...$$
$$+ a_{p-2}(\zeta^{p-2} - \zeta^{p-1})$$

By the linearity of the trace and our above calculations we find that $Tr_{K/\square}(\alpha(1-\zeta)) = pa_0$ We also have
$Tr_{K/\square}(\alpha(1-\zeta)) \in p\square$, so $a_0 \in \square$ Next consider the algebraic integer
$(\alpha - a_0)\zeta^{-1} = a_1 + a_2\zeta + ... + a_{p-2}\zeta^{p-3}$; This is an algebraic integer since $\zeta^{-1} = \zeta^{p-1}$ is. The same argument as above shows that $a_1 \in \square$, and continuing in this way we find that all of the a_i are in $\square$. This completes the proof.

Example 1.4   Let $K = \square$, then the local ring $\square_{(p)}$ is simply the subring of $\square$ of rational numbers with denominator relatively prime to p. Note that this ring $\square_{(p)}$ is not the ring $\square_p$ of p-adic integers; to get $\square_p$ one must complete $\square_{(p)}$. The usefulness of $O_{K,p}$ comes from the fact that it has a particularly simple ideal structure. Let a be any proper ideal of $O_{K,p}$ and consider the ideal $a \cap O_K$ of O_K. We claim that $a = (a \cap O_K)O_{K,p}$; That is, that a is generated by the elements of a in $a \cap O_K$. It is clear from the definition of an ideal that $a \supseteq (a \cap O_K)O_{K,p}$. To prove the other inclusion, let α be any element of a. Then we can write $\alpha = \beta/\gamma$ where $\beta \in O_K$ and $\gamma \notin p$. In particular, $\beta \in a$ (since $\beta/\gamma \in a$ and a is an ideal), so $\beta \in O_K$ and $\gamma \notin p$. so $\beta \in a \cap O_K$. Since $1/\gamma \in O_{K,p}$, this implies that $\alpha = \beta/\gamma \in (a \cap O_K)O_{K,p}$, as claimed. We can use this fact to determine all of the ideals of $O_{K,p}$. Let a be any ideal of $O_{K,p}$ and consider the ideal factorization of $a \cap O_K$ in O_K. write it as $a \cap O_K = p^n b$ For some n and some ideal b, relatively prime to p. we claim first that $bO_{K,p} = O_{K,p}$. We now find that

$$a = (a \cap O_K)O_{K,p} = p^n bO_{K,p} = p^n O_{K,p}$$

Since $bO_{K,p}$. Thus every ideal of $O_{K,p}$ has the form $p^n O_{K,p}$ for some n; it follows immediately that $O_{K,p}$ is noetherian. It is also now clear that $p^n O_{K,p}$ is the unique non-zero prime ideal in $O_{K,p}$. Furthermore, the inclusion $O_K \mapsto O_{K,p}/pO_{K,p}$ Since $pO_{K,p} \cap O_K = p$, this map is also surjection, since the residue class of $\alpha/\beta \in O_{K,p}$ (with $\alpha \in O_K$ and $\beta \notin p$) is the image of $\alpha\beta^{-1}$ in $O_{K/p}$, which makes sense since β is invertible in $O_{K/p}$. Thus the map is an isomorphism. In particular, it is now abundantly clear that every non-zero prime ideal of $O_{K,p}$ is maximal.   To show that $O_{K,p}$ is a Dedekind domain, it remains to show that it is integrally closed in K. So let $\gamma \in K$ be a root of a polynomial with coefficients in $O_{K,p}$; write this polynomial as

$$x^m + \frac{\alpha_{m-1}}{\beta_{m-1}}x^{m-1} + ... + \frac{\alpha_0}{\beta_0}$$ With $\alpha_i \in O_K$ and $\beta_i \in O_{K-p}$. Set $\beta = \beta_0\beta_1...\beta_{m-1}$. Multiplying by β^m we find that $\beta\gamma$ is the root of a monic polynomial with coefficients in O_K. Thus $\beta\gamma \in O_K$; since $\beta \notin p$, we have $\beta\gamma/\beta = \gamma \in O_{K,p}$. Thus $O_{K,p}$ is integrally close in K.

COROLLARY 1.2.   Let K be a number field of degree n and let α be in O_K then
$$N_{K/\square}'(\alpha O_K) = \left| N_{K/\square}(\alpha) \right|$$
PROOF. We assume a bit more Galois theory than usual for this proof. Assume first that $K/\square$ is Galois. Let σ be an element of $Gal(K/\square)$. It is clear that $\sigma(O_K)/\sigma(\alpha) \cong O_{K/\alpha}$; since $\sigma(O_K) = O_K$, this shows that $N_{K/\square}'(\sigma(\alpha)O_K) = N_{K/\square}'(\alpha O_K)$. Taking the product over all $\sigma \in Gal(K/\square)$, we have $N_{K/\square}'(N_{K/\square}(\alpha)O_K) = N_{K/\square}'(\alpha O_K)^n$ Since $N_{K/\square}(\alpha)$ is a rational integer and O_K is a free $\square$-module of rank n,
$O_K/N_{K/\square}(\alpha)O_K$ Will have order $N_{K/\square}(\alpha)^n$; therefore
$$N_{K/\square}'(N_{K/\square}(\alpha)O_K) = N_{K/\square}(\alpha O_K)^n$$

This completes the proof. In the general case, let L be the Galois closure of K and set $[L:K]=m$.

V. CHALLENGES TO THE QUALITY OF POWER AND COMMUNICATIONS FOR THE SMART GRID

The following elaborate on a series of essential attributes for the smart grid, as shown in Figure 5. As a matter of fact, they are interconnected in a very close relationship as cause-effect among one another. Each of them should be well considered for the progressive strategies and designs in both power and communications fields to support the smart grid development.

A. Reliability and Stability

Reliability has a durability feature. In general, it interprets the operational health and degree of volatility of the entire system. It further presents the state of high consistency, repeatability, and dependability that the smart grid will maintain in accordance with effective measurements and assessments. The stability of a system determines the level of reliability for which the system is qualified. Meanwhile, reliability is dependent upon the achievement of other contributing and decisive factors, described in the following subsections. The smart grid must guarantee voltage and current stability, mitigate peak demand and load variability with implementation of DG and energy storage over wide areas, and preclude a variety of incidents.

B. Measurability and Controllability

Service interruption and faults are serious and possible to happen. It is important for them to be measurable and controllable in such a way that deliberate evaluations and assessments can take place. The smart grid is capable of identifying and correcting disruptive operations through dynamic measurements and control monitoring in real time. In the meantime, some degrees of observability and transparency should be obtained in order to efficiently analyze, manage, as well as predict and respond to varying network conditions. A wealth of data information, which considerably makes the grid smart, must also be measurable, observable, and manageable.

C. Flexibility and Scalability

The grid is moving from a centralized infrastructure to multiple decentralized MGs. Scalability of the electric power system has to be well determined. By means of islanding, MGs attempt to incorporate DG and energy storage for contributing energy to utilities as well as to local served areas during times of peak demand. The island functionality [61] introduces a concept of a giant smart grid composed of multiple micro smart grids. Each local grid is able to operate autonomously with respect to DSM, quality and reliability model, fault management, and security management. While fulfilling the scalability goal, flexibility allows the smart grid to provide multiple redundant alternate routes for power and data to flow, as well as supply options for feasible control and operation when needed. Flexibility may also apply to a set of standards exercised in the smart grid such that they should be available and upgradeable throughout the world including ANSI, IEC, PLC, wireless M-Bus, and ZigBee.

D. vailability

The availability of power and communications is essential upon consumers' requests for energy and information. It relies upon the availability of data exchanged in the network. A high degree of resource availability is required especially when dealing with the latency and security issues. For examples, a latency of few tens of milliseconds should be achieved in line protection and control systems; a DoS attack can aggravate network performance causing servers or services temporarily unavailable. The redundancy measure might be a foreseeable resolution to the problem. However, its effectiveness will depend upon how the system should be designed while avoiding the consequent costs of high network complexity as well as the scalability matter.

E. Resiliency

The degree of resiliency determines how reliable the smart grid actually is when incidents happen. Especially from the safety and security perspectives, resiliency represents the capability to restore and recover from various disruptions or malfunctions through the robust fast-response process. The vulnerable electric components in the grid are likely transmission lines and stations, large power plants, and nuclear plants with leakage. Multiple contingency plans are required to address any failures or destruction caused by natural disasters, deliberate attacks, or malicious activities.

F. Smart Failover/Fail-back Mechanism

The shortcoming of this failover/fail-back mechanism described previous section is that it is not graceful when it comes to job management. Whenever a failover to backup occurs, all the jobs running and in the job queue are lost and must be resubmit at startup of both backup- and primary-server. This situation is particularly aggravated when jobs must be submitted numerous times due to multiple fail/recovery restarts. Furthermore, those jobs still running will have to discard their "preliminary" results and start over in the job queue. Timely check pointing is one obvious solution but not many schedulers come with an automatic checkpoint/restart facility. The OpenPBS scheduler offers check-pointing facility for the SGI and IRIX

platforms but not for others. Thus, this situation must be properly handled and to do so, we introduce the "smart failover/fail-back mechanism" in HA-OSCAR. This mechanism periodically saves job states in the queue and rsyncs those states to the backupserver machine. When the backup-server is called to action, it will then start jobs from its last saved job state. Thus, with the exception of any running job, we are able to guarantee the processing of jobs in the queue and in queue order. This feature also alleviates the user from the task of resubmitting all those jobs that were sitting in the queue at the time of failure. If the jobs were submitted locally, by non-Globus mechanisms, their output and error logs can be retrieved from files specified in the script file input to qsub. When the jobs are submitted through the grid, the Globus gatekeeper invokes a job manager which in turn invokes the scheduler specified (GT 2.4 specific). The job manager co-ordinates the transferring of output and error logs to the client who submitted the jobs. Whenever we failover all connections are lost and as the jobs are restarted on the backup, the job manager is not in place to collect the output and error log from the jobs. This scenario can be handled by updating the saved job state files to redirect the output and error logs to files Proceedings of the 19th International Symposium on High Performance Computing Systems and Applications having names starting with job ids. In this way the client can fetch those files to view later as the output and error logs are based on the job ids of their jobs. This is an area of future research. The situation gets a bit complex when dealing with fail-back, as the standby immediately gives up its IP to the primary and there is no chance of transferring the jobs states back to the primary. This situation can be handled if we have a shared storage medium between the Primary and standby nodes. This too is an area of future research. There is presently a prototype implementation supporting the smart failover mechanism in our lab that has not yet been publicly released.

G. Automated Grid Installation Package

As mentioned earlier, the installation and configuration of Grid Tools and packages becomes challenging for novice users. This installation and configuration should be automated and efforts have been initiated in that direction. ROCKS provides a package called "Grid-Roll" which installs the Globus toolkit during setup thus alleviating the user of many setup and configuration tasks. Similarly, as part of the OSCAR group, we are developing an automated Globus toolkit and accessories installation package. This OSCAR grid package will enable an OSCAR cluster software stack installation for both the standard release as well as the HA-OSCAR environment. Before the installation of the Globus toolkit we need to verify that the support software is in place and the environment is properly

setup. Even after the installation we need to manually tweak some files such as adding the Globus library path to /etc/ld.so.conf, creating xinetd.d files for gatekeeper and grid-ftp services, installation of support software for interfacing with local schedulers such as PBS, Condor and much more. All these steps, if automated, shall help reducing the planned downtime of the system for such installations and configurations on OSCAR clusters. We have created prototype version of this installation process using the OSCAR packaging framework.

H. Implementation and Experimental Results

Figure 4 shows the setup used in this experiment. The site-manager is a cluster head node based on the OSCAR installation. We have installed Globus Toolkit version 2 (GT2.4) and run grid services there as we were dealing with pre web services only. We also installed the client utilities of GT2.4 on a client machine in the grid. Later we enabled the site-manager with HAOSCAR to create a standby node, an active redundancy, which was the exact clone of the site-manager. The HA-OSCAR monitoring service monitored the gatekeeper and gridFTP services on the server every 3 seconds. Every 3 seconds the script tried to establish telnet connections to the port numbers on which the Globus gatekeeper and gridFTP accept connections. If the connection was established we assumed that the service was alive and working. In addition, critical services such as xinetd, SGE, Condor, PBS and NFS were also being monitored with the corresponding policy for recovery from failure. The standby node also monitors the primary server every 3 seconds for availability.

Fig. 4 Setup used to perform the experiment

During the first set of experiments, we simulated service outages by killing the Xinetd daemon and observed whether a failure of gatekeeper or gridFTP services (as they are started on demand by Xinetd daemon) was detected and then restarted by the HA-OSCAR recovery mechanism. Normal operation was validated after ensuring that client was able to run "globus-job-run" using both the fork job manager as well as PBS jobmanager and "globus-url-copy" commands successfully. We ran a MPI job using the PBS job manager for validating the server client connectivity. The "globus-url-copy" command involves the use of the gridFTP server for transferring files from one

VI. ANALYSIS OF DATA CHARACTERISTICS
A. Diversity of Data Sources

The Smart Grid's intelligence and adaptiveness depends on the ability to acquire and integrate diverse information that help perform accurate load forecasting and curtailment by utilities and provide rich services to customers. A Smart

Grid utility uses both direct power systems information and information that indirectly helps forecast, correlate and control power usage. Direct information sources include consumer smart meters that transmit power usage and smart appliances data, sensors at transformers and distribution stations, and customer information systems used for billing. Indirect sources are historical, current and forecast weather from NOAA, social network and schedule information shared by consumers for load prediction, studying consumer behavior on the utility's website, and mobile applications that may send consumer location information and receive load curtailment response. The conceptual diversity present in the Smart Grid system gives rise to a wider range of information from multiple sources that need to be secured and controlled according to policies defined by the data owners. These data sources include information that is both public and private, with ownership belonging to the different user roles introduced before. Increasing information flows raises the chance that personally identifiable information will be passed which, if not handled carefully, can lead to violation of an individual's privacy. Cloud platforms need to support secure data acquisition from different information sources. While public Clouds are naturally suited for scaling out and processing millions of user requests, the diversity of information also requires diverse storage services that can enforce security and privacy policies. The policies themselves can be complex and varied, given the number of different information sources such as consumers, public agencies, online service providers and prior utility data.

B. Data Size and Temporal Granularities

Smart Grid utilities need to handle data at extreme scales of data size. At one end, HAN systems can report fine-grained usage of smart appliances, on the order of bytes/kilobytes to the utility through the smart meter. At the other end, this data accumulated from millions of consumers over years can grow to petabytes (PB) in size, and form a data mining corpus to detect load patterns and test response scenarios. The size of data collected may vary continuously as adaptive demand-response algorithms control smart meter data collection rates, and add or drop information sources [11]. Privacy policies and security infrastructure has to efficiently and effectively support such diverse information sizes. The frequency of data generation and its timeliness of use in Smart Grids also differs from traditional power grids.

VII. BUILDING A UNIFIED SOFTWARE SERVICES INTERFACE

The Software Service Interface (SSI, or Software Systems Interface) is a middleware between applications, fundamental smart grid services, and third-party services. Typically, SSI is implemented as a bus where different applications and services can post and exchange messages. Examples of applications include intelligent VAR control systems, demand response programs, and substation automation. Fundamental application services include load flow, state estimation, and load forecasting, among others. Because SSI is a comprehensive architecture, SSI is an enabler for communications between smart grid applications and services. The ambitious goal set forth by SSI proponents is to achieve a universal communications platform between applications and basic services. SSI cannot be a product or an application by itself.

The need to know about the types of data transmitted through the SSI bus (expected inputs and outputs in the appropriate format) may lead to over-engineering and distract a software developer from the initial implementation of the communications platform.

A. Acknowledging Needs for SSI

Depending on individual situations, SSI may prove superior to other architectures:
- When multiple services and applications need to communicate with each other,
- With large DMS/EMS applications deployments that utilize multiple services and multiple applications,
- When modules need to listen to all system events in a way that is similar to the IEC 61850 substation event bus.

B. Software Services Communications Standards

The IEC 61850 [8] standard defines how substation equipment should communicate among each other. CIM [9] standardizes how models are coded to facilitate exchange of model data between GIS, DMS, and EMS. Existing communications standards include the DNP protocol in substations, IEEE 802.11 [10] (ethernet for wireless devices), the Zigbee protocol for communications between energy devices in a wireless home area network (IEEE 802.15.4) [11], and, more recently, 3G and 4G cell phone standards.

C. Development and Testing Challenge

Because each standard defines its own requirements, an Agile development process to support all the standards must implement and test requirements in a priority order that allows customer functionality to be implemented. Following such a gradual approach increases the quality and robustness of each supported aspect of a standard. Because SSI aims at comprehensiveness, the testing of individual functionality is more challenging than with simple architectures. Quality is more difficult to assess with more complicated services. However,

using an approach such as Agile helps determine the development steps to gradually implement SSI depending on business needs.

VIII. REAL TIME REQUIREMENTS

Smart grid software must process models and measurements so that utilities and consumers can have access to power data in real time. Smart grid technology is expected to optimize large distribution networks within minutes and obtain the system state several times per minute. In this section, we examine the various implications for the development of real-time smart-grid software services.

A. Speed Requirements

Recent distribution automation applications must process circuits in real time to provide operating recommendations. Distribution optimization applications scan feeder data and measurements for improvements and should process several million nodes within minutes. Such desired processing speeds impose strict time requirements to the distribution load flow, state estimation, and optimization processes. Load flow may be needed a few dozen times per second, and state estimation up to ten times per second. Architectural overhead increases the total processing time depending on architecture complexity. The total processing time is tied to architectural choices as well as the performance of the individual components of the chosen architecture.

B. Programming Platforms

Programming platform come with tradeoffs between performance control and ease of development. Platforms such as Java can be mastered quickly; however, automatic garbage collection is often times a handicap in real-time programming [8, 9], especially when programmers have no control on the duration of a garbage collection cycle. As an alternative, the Ruby platform offers control over garbage collection. In contrast, C/C++ is usually fastest as it compiles directly to CPU instructions. Granularity of control to the developers is greater than with Java or similar languages; however, memory management is the full responsibility of the developer.

C. Software Complexity

Most of the time, software complexity and layers such as security layers reduce real-time performance. Simple software service architectures are easier to fine-tune for performance than large architectures such as SSI. The different software layers in large architectures increase overhead processing without adding value, but are useful if disparate communication protocols exist between software services.

D. Communications Delays/Lags

Communications delays and lags caused by long communications lines must be accounted for. Time stamps may be applied to mark time-sensitive data. Also, exchanging megabytes of data several times a second between servers may raise questions about where the data should be handled, as fewer batches of data exchanged also mean less time spent encoding or decoding the data.

E. Server Requirements

Communications in real-time services should be limited to the originating and destination servers. Compared to the VIP architecture, and depending on implementation, additional redundant servers may be required to dispatch the messages posted on the SSI bus while maintaining service continuity in the event of a server failure.

F. Security and Data Integrity

Utilities take security seriously, given the potential for a hacker to penetrate utility servers and issue commands that could take an electric network out of control. The time needed to authenticate, process the security layer, and to encrypt/decrypt data with a strong cipher must be accounted for in an application.

IX. DEMAND RESPONSE AND END-USERS PREFERENCES FOR EFFICIENT CONSUMPTION OF ENERGY

An effective approach for achieving demand response requires techniques at the consumption level too. This is done by having an intelligent framework at the consumption side. For example, at home level, input is taken from the grid and, depending on various underlying factors, the framework assists the consumer to achieve demand response. There are no related works in the literature in this regard but the importance of such work has been discussed by Hopper et al. [27] who state that there is a role for targeted technical assistance programs to help customers to develop more sophisticated price response strategies as shown in Figure 3. There is a utility provider which sends price signals to end-users and receives the consumption information by means of a smart meter and wireless communication of the sensors (smart appliances).

The consumer mutually receives some information from the utility provider about consumption profile and price signals from various portals. It is expected that, by receiving consumption information, the consumers will change their consumption behaviour in order to mitigate cost and save on their power bill. However, in a dynamic pricing system, the consumers have no way of knowing whether their decision to modify their energy consumption is effective and efficient. This is

overcome by adding intelligence at each home level. In this paper, we will develop a model by which such intelligence is added at each home level on a continuous basis by which demand response is achieved. We will propose our model in the next section.

A. Authors and Affiliations

Dr Akash Singh is working with IBM Corporation as an IT Architect and has been designing Mission Critical System and Service Solutions; He has published papers in IEEE and other International Conferences and Journals.

He joined IBM in Jul 2003 as a IT Architect which conducts research and design of High Performance Smart Grid Services and Systems and design mission critical architecture for High Performance Computing Platform and Computational Intelligence and High Speed Communication systems. He is a member of IEEE (Institute for Electrical and Electronics Engineers), the AAAI (Association for the Advancement of Artificial Intelligence) and the AACR (American Association for Cancer Research). He is the recipient of numerous awards from World Congress in Computer Science, Computer Engineering and Applied Computing 2010, 2011, and IP Multimedia System 2008 and Billing and Roaming 2008. He is active research in the field of Artificial Intelligence and advancement in Medical Systems. He is in Industry for 18 Years where he performed various role to provide the Leadership in Information Technology and Cutting edge Technology.

REFERENCES

[1] Dynamics and Control of Large Electric Power Systems. Ilic, M. and Zaborszky, J. John Wiley & Sons, Inc. © 2000, p. 756.

[2] Modeling and Evaluation of Intrusion Tolerant Systems Based on Dynamic Diversity Backups. Meng, K. et al. Proceedings of the 2009 International Symposium on Information Processing (ISIP'09). Huangshan, P. R. China, August 21-23, 2009, pp. 101–104

[3] Characterizing Intrusion Tolerant Systems Using A State Transition Model. Gong, F. et al., April 24, 2010.

[4] Energy Assurance Daily, September 27, 2007. U.S. Department of Energy, Office of Electricity Delivery and Energy Reliability, Infrastructure Security and Energy Restoration Division. April 25, 2010.

[5] CENTIBOTS Large Scale Robot Teams. Konoledge, Kurt et al. Artificial Intelligence Center, SRI International, Menlo Park, CA 2003.

[6] Handling Communication Restrictions and Team Formation in Congestion Games, Agogino, A. and Tumer, K. Journal of Autonomous Agents and Multi Agent Systems, 13(1):97–115, 2006.

[7] Robotics and Autonomous Systems Research, School of Mechanical, Industrial and Manufacturing Engineering, College of Engineering, Oregon State University

[8] D. Dietrich, D. Bruckner, G. Zucker, and P. Palensky, "Communication and computation in buildings: A short introduction and overview," *IEEE Trans. Ind. Electron.*, vol. 57, no. 11, pp. 3577–3584, Nov. 2010.

[9] V. C. Gungor and F. C. Lambert, "A survey on communication networks for electric system automation," *Comput. Networks*, vol. 50, pp. 877–897, May 2006.

[10] S. Paudyal, C. Canizares, and K. Bhattacharya, "Optimal operation of distribution feeders in smart grids," *IEEE Trans. Ind. Electron.*, vol. 58, no. 10, pp. 4495–4503, Oct. 2011.

[11] D. M. Laverty, D. J. Morrow, R. Best, and P. A. Crossley, "Telecommunications for smart grid: Backhaul solutions for the distribution network," in *Proc. IEEE Power and Energy Society General Meeting*, Jul. 25–29, 2010, pp. 1–6.

[12] L. Wenpeng, D. Sharp, and S. Lancashire, "Smart grid communication network capacity planning for power utilities," in *Proc. IEEE PES, Transmission Distrib. Conf. Expo.*, Apr. 19–22, 2010, pp. 1–4.

[13] Y. Peizhong, A. Iwayemi, and C. Zhou, "Developing ZigBee deployment guideline under WiFi interference for smart grid applications," *IEEE Trans. Smart Grid*, vol. 2, no. 1, pp. 110–120, Mar. 2011.

[14] C. Gezer and C. Buratti, "A ZigBee smart energy implementation for energy efficient buildings," in *Proc. IEEE 73rd Veh. Technol. Conf. (VTC Spring)*, May 15–18, 2011, pp. 1–5.

[15] R. P. Lewis, P. Igic, and Z. Zhongfu, "Assessment of communication methods for smart electricity metering in the U.K.," in *Proc. IEEE PES/IAS Conf. Sustainable Alternative Energy (SAE)*, Sep. 2009, pp. 1–4.

[16] A. Yarali, "Wireless mesh networking technology for commercial and industrial customers," in *Proc. Elect. Comput. Eng., CCECE,* May 1–4, 2008, pp. 000047–000052.

[17] M. Y. Zhai, "Transmission characteristics of low-voltage distribution networks in China under the smart grids environment," *IEEE Trans. Power Delivery*, vol. 26, no. 1, pp. 173–180, Jan. 2011.

[18] V. Paruchuri, A. Durresi, and M. Ramesh, "Securing powerline communications," in

Proc. IEEE Int. Symp. Power Line Commun. Appl., (ISPLC), Apr. 2–4, 2008, pp. 64–69.

[19] Q.Yang, J. A. Barria, and T. C. Green, "Communication infrastructures for distributed control of power distribution networks," *IEEE Trans. Ind. Inform.*, vol. 7, no. 2, pp. 316–327, May 2011.

[20] T. Sauter and M. Lobashov, "End-to-end communication architecture for smart grids," *IEEE Trans. Ind. Electron.*, vol. 58, no. 4, pp. 1218–1228, Apr. 2011.

[21] K. Moslehi and R. Kumar, "Smart grid—A reliability perspective," *Innovative Smart Grid Technologies (ISGT)*, pp. 1–8, Jan. 19–21, 2010.

[22] Southern Company Services, Inc., "Comments request for information on smart grid communications requirements," Jul. 2010

[23] R. Bo and F. Li, "Probabilistic LMP forecasting considering load uncertainty," *IEEE Trans. Power Syst.*, vol. 24, pp. 1279–1289, Aug. 2009.

[24] *Power Line Communications*, H. Ferreira, L. Lampe, J. Newbury, and T. Swart (Editors), Eds. New York: Wiley, 2010.

[25] G. Bumiller, "Single frequency network technology for fast ad hoc communication networks over power lines," WiKu-Wissenschaftsverlag Dr. Stein 2010.

[31] G. Bumiller, L. Lampe, and H. Hrasnica, "Power line communications for large-scale control and automation systems," *IEEE*

Commun. Mag., vol. 48, no. 4, pp. 106–113, Apr. 2010.

[32] M. Biagi and L. Lampe, "Location assisted routing techniques for power line communication in smart grids," in *Proc. IEEE Int. Conf. Smart Grid Commun.*, 2010, pp. 274–278.

[33] J. Sanchez, P. Ruiz, and R. Marin-Perez, "Beacon-less geographic routing made partical: Challenges, design guidelines and protocols," *IEEE Commun. Mag.*, vol. 47, no. 8, pp. 85–91, Aug. 2009.

[34] N. Bressan, L. Bazzaco, N. Bui, P. Casari, L. Vangelista, and M. Zorzi, "The deployment of a smart monitoring system using wireless sensors and actuators networks," in *Proc. IEEE Int. Conf. Smart Grid Commun. (SmartGridComm)*, 2010, pp. 49–54.

[35] S. Dawson-Haggerty, A. Tavakoli, and D. Culler, "Hydro: A hybrid routing protocol for low-power and lossy networks," in *Proc. IEEE Int. Conf. Smart Grid Commun. (SmartGridComm)*, 2010, pp. 268–273.

[36] S. Goldfisher and S. J. Tanabe, "IEEE 1901 access system: An overview of its uniqueness and motivation," *IEEE Commun. Mag.*, vol. 48, no. 10, pp. 150–157, Oct. 2010.

[37] V. C. Gungor, D. Sahin, T. Kocak, and S. Ergüt, "Smart grid communications and networking," Türk Telekom, Tech. Rep. 11316-01, Apr 2011.